

NASA CR-132851

FINAL REPORT

ADVANCED ON-BOARD PROCESSOR (A.O.P.)

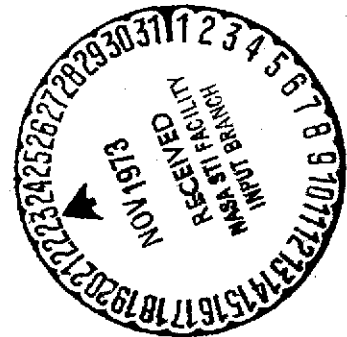
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Prepared for

NATIONAL AERONAUTICS AND SPACE ADMINISTRATION

Goddard Space Flight Center

Greenbelt, Maryland



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(AOP)

June 15, 1973

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2.0 INTRODUCTION AND SUMMARY

The Westinghouse Electric Corporation has delivered the Advanced On-Board Processor (AOP) to NASA/GSFC. This computer evolved from the earlier On-Board Processor (OBP) which is now performing perfectly in flight aboard the OAO-C satellite. Both the OBP and the AOP were designed, fabricated and tested by a very closely coordinated program between Westinghouse and NASA/GSFC..

The AOP uses Large Scale Integration throughout and is the most advanced space qualified computer of its class in existence today. It was designed to satisfy most spacecraft requirements which are anticipated over the next several years.

The AOP design utilizes Custom Metallized Multigate Arrays (CMMA) which have been designed specifically for this computer. This approach provides the most efficient use of circuits, reduces volume, weight, assembly costs and provides for a significant increase in reliability by the significant reduction in conventional circuit interconnections.

The required 69 CMMA packages are assembled on a single multilayer printed circuit board which together with associated connectors constitutes the complete AOP. This approach also reduces conventional interconnections thus further reducing weight, volume and assembly costs.

This approach together with a CMMA-AOP breadboard (fabricated during the program) permits the computer to be completely checked out prior to assembly. Thus minimizing last minute problems which inevitably occur.

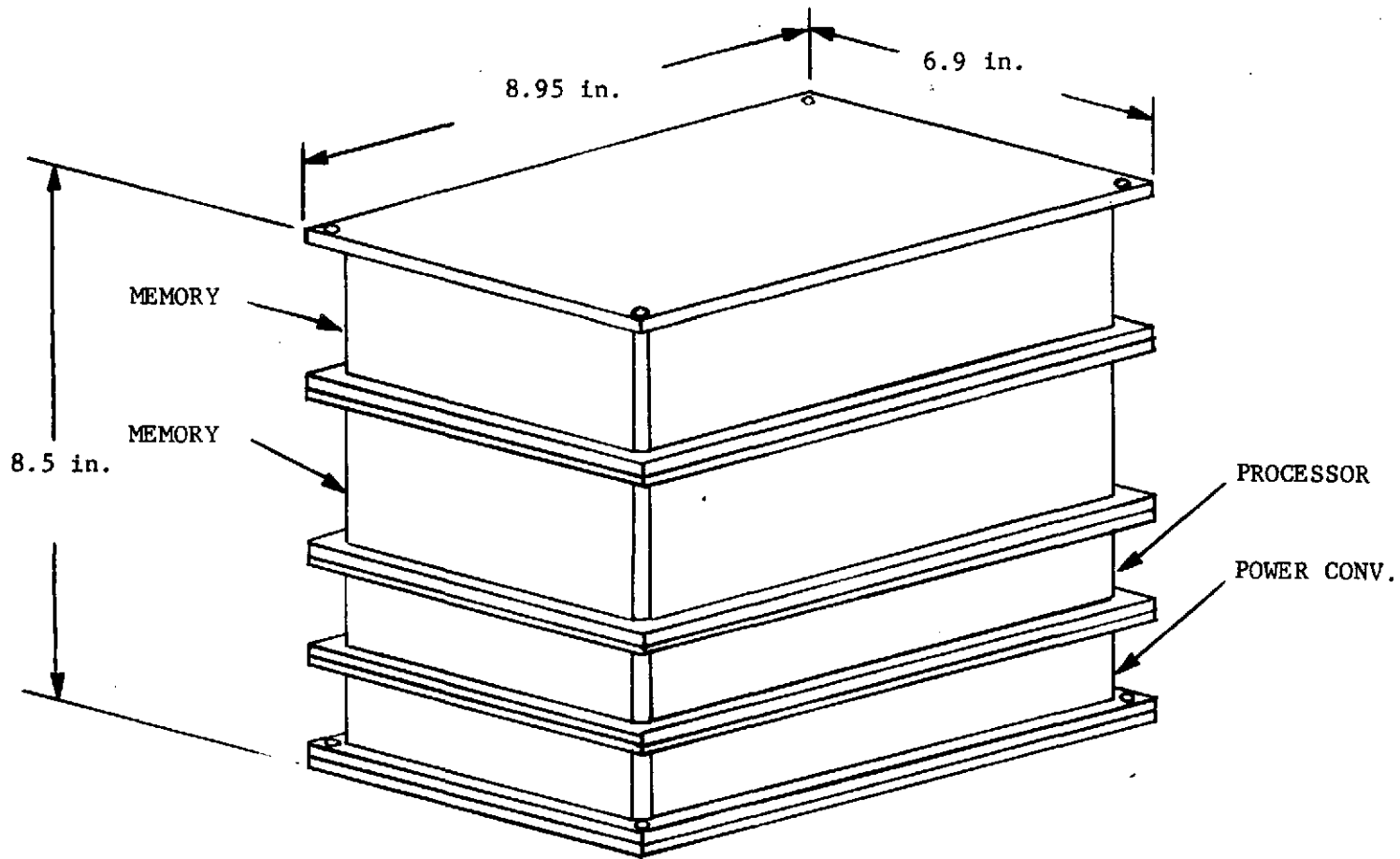
The important characteristics of the AOP are shown in table 2-1 and the configuration of the computer, special I/O, power supply and 8K of memory is shown in figure 2-1. The AOP is defined in detail in the attached appendix. Logic diagrams have been delivered to GSFC and are available at that facility.



1. WORD LENGTH - 18 Bits
2. INSTRUCTION WORD - 18 Bits
3. DATA FLOW - Parallel
4. DATA TYPE - Fixed Point - Fraction - 2's Complement
5. NO. OF INSTRUCTIONS - 55
6. ADD TIME - 5.0 Microseconds
7. MULTIPLY TIME - 38 Microseconds
8. DIVIDE TIME - 75 Microseconds
9. CLOCK RATE - 800 KHz
10. NO. OF INDEX REGISTERS - One
11. OPERATIONS PER SECOND - 200K
12. INDIRECT ADDRESSING - Load and Store
13. DOUBLE LENGTH ACCUMULATOR
14. TECHNOLOGY - TTL-LSI
15. SIZE - 112 cu. in. (including all I/O)
16. WEIGHT - 3 pounds (including all I/O)
17. POWER CONSUMPTION - Watts (CPU-Fixed I/O)
18. INTERRUPTS - 16 multilevel interrupts with program lockout control and interrupt override feature.
19. DIRECT MEMORY ACCESS (DMA) - Up to 16 devices time share one channel maximum I/O rate - 66K words per second.
20. COMMAND LOAD AND DUMP-GROUND COMMAND - Control for direct memory load or dump maximum load word rate - 300K words per second.
21. MEMORY CAPABILITY - 4K modules expandable to 64K (each module incorporates cycle by cycle power switching).
22. Direct Addressing of 4096 words in any page (addressing of all 64K of memory is accomplished by indexing or paging).
23. MEMORY WRITE PROTECTION - Storage area assigned in 128 word blocks.
24. Busing Organization for parallel unpowered spare processor and/or memory modules.

TABLE 2-1

AOP CHARACTERISTICS



5

FIGURE 2-1 TYPICAL ACP STACKED CONFIGURATION



Aerospace Division



The AOP was designed especially to satisfy the requirements of all classes of spacecraft including small satellites. In addition to practical experience gained on the OAO-C with the OBP other systems requirements were also studied before the design goals for the AOP were finalized. As a result the following major design features have been achieved.

. Low Power Dissipation

When operating at 100% duty cycle the power consumption is 5.0 watts including the I/O section.

. Low Volume and Weight

The volume is approximately 112 cu. inches and the weight is 3 lbs. including the I/O section and the special I/O section.

. High Speed Arithmetic and Logic Capability

The AOP is approximately twice as fast as the OBP.

. Flexible Input Output Design

Program independent data transfers are accomplished through the use of 16 buffered I/O channels operating in a cycle steal mode, sharing common Direct Memory Access (DMA) hardware.

. Interrupt Philosophy

Sixteen individually armed priority interrupts allow asynchronous spacecraft events to gain access to computer operation at event dependent intervals.

. Modularity

The nominal AOP system consists of two "double ported" processor modules and a minimum of two memory modules connected by two data buses. Each memory is dedicated to a single data bus. The modules along with the power supply module are packaged to provide a mechanically integrated package. The size is expandable depending on system requirements. The



mechanical arrangement is shown in figure 2-1 and the data bus diagram is shown in figure 2-2.

• Reliability

Reliability at the component level was **achieved** by the following steps:

1. Simple designs using minimum components.
2. Proven process for LSI Arrays (TTL).
3. Proven metalization process for CMMA.
4. High reliable screening during CMMA production.

System reliability depends on the mission constraints and requires either a simplex system or dual simplex system shown in figure 2-3 (a), or a single duplex system shown in figure 2-3 (b). The AOP uses the single duplex system of figure 2-3 (b).

• Support Software

Support software consists of an assembler, loader and simulator. These programs are embedded in a flexible control language. The entire system is link structured Fortran program with minimal use of assembly language. Minimum configuration is an SDS 920.

• Spacecraft Applications

Interconnection with remote multiplexers can provide the computer with access to all onboard data. The computer can process and/or format data in a programmable manner which provides great flexibility to any mission. Digital to Analog converters can provide analog signals for control loops.

Table 2-2 is an attempt to summarize areas in which the computer can contribute or enhance operations capability on future satellites. It should be emphasized that this is an incomplete list and is growing

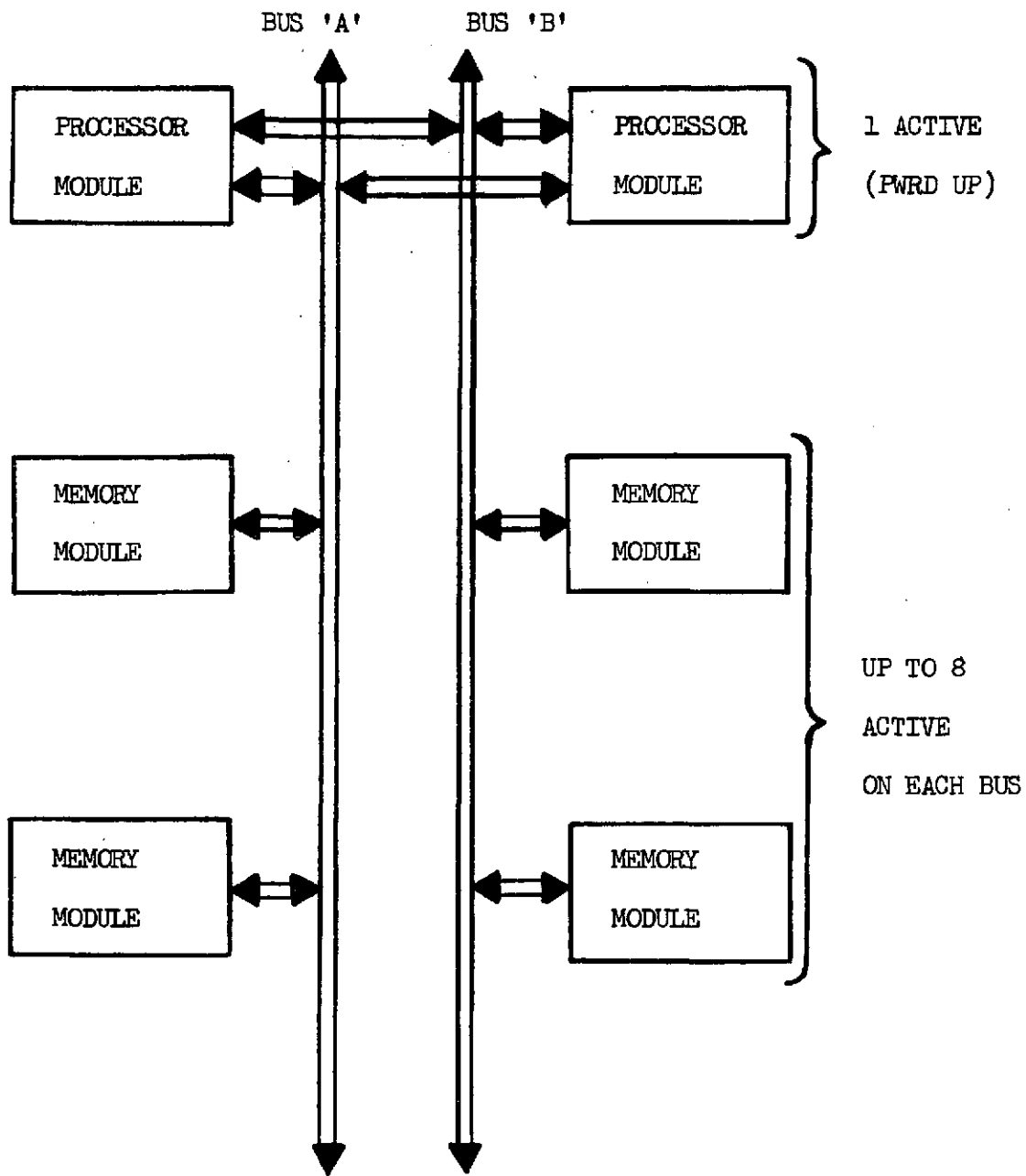
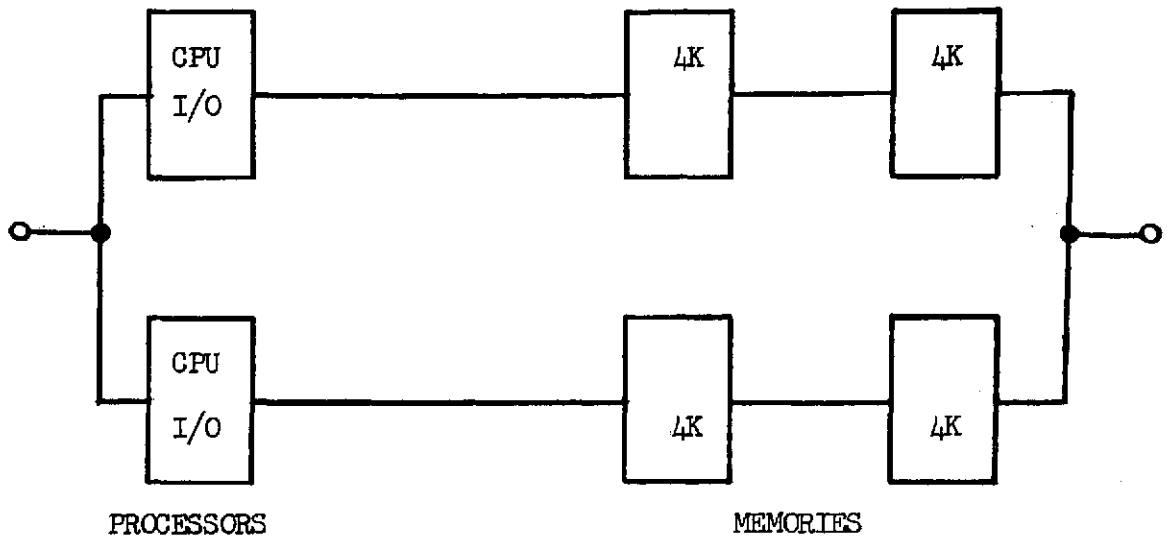
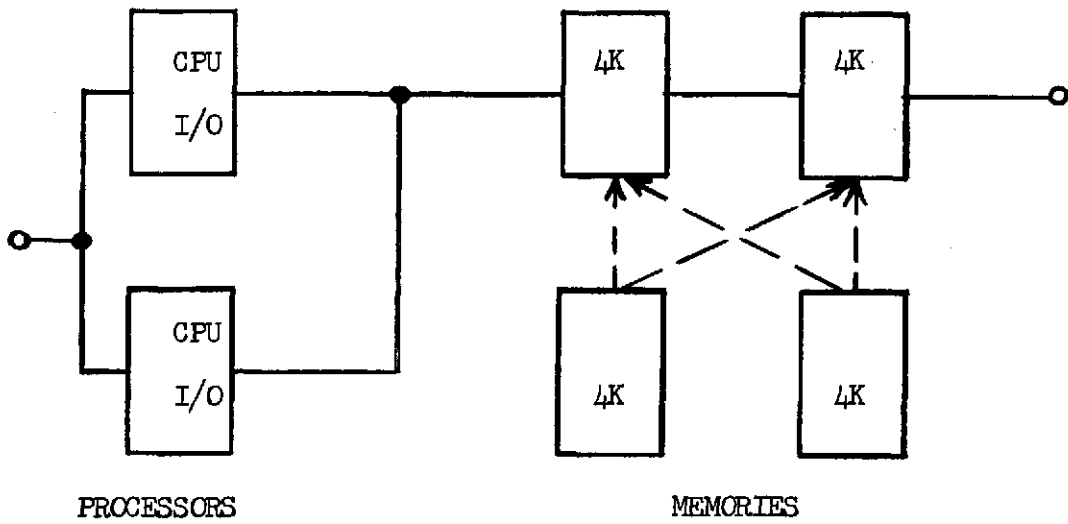


FIGURE 2-2 AOP SYSTEM BLOCK DIAGRAM



(a) Dual Simplex Design



ANY 2 OF 4 MODULES MUST OPERATE
6 POSSIBLE COMBINATIONS

(b) Single Duplex Design (AOP)

FIGURE 2-3 RELIABILITY MODELS



Command Handling

1. Storage for Delayed Execution
 - a. Long Term Schedule
 - b. Special "Scratch Pad" Requests
2. Execution of Stored Sequences
3. Execution of Data Dependent Commands

Data Handling

1. Format Control
2. Data Compression
3. Status Summary
4. Operations Summary
5. Limit Checking

Spacecraft Operation

1. Emergency Control Sequences
2. Failure Workarounds
3. Thermal Control
4. Power Scheduling
5. Array/Regulator Control
6. Attitude Control
 - a. Stabilization
 - b. Pointing/Maneuvering
 - c. Momentum Management
7. System Test and Diagnosis

Experiment Operation

1. Event/Data Dependent Control/Commands
2. Routine Experiment Mode Control and Operation Scheduling
3. Test and Diagnosis
4. Sensor Stimulation and/or Calibration
5. Data Processing

TABLE 2-2

LIST OF ON-BOARD COMPUTER FUNCTIONS



everyday as new experience is gained on present day computerized spacecraft.

3.0 AOP BREADBOARD

An AOP Breadboard was designed using small scale integrated circuits (SSI) in DIP packages and discrete circuits to verify the functional operation of the AOP logic.

The logic of the breadboard was partitioned in groupings representative of the expected CMMA devices. Off the shelf Augat wire wrap boards and a wire wrap mother board were used, each containing logic equivalent to three CMMA devices. The completed breadboard required 23 wire wrap boards and two mother boards of 13 slots each.

3.1 Circuits

The circuits used in the breadboard were a mix of TTL and DTL with low powered 54L series being used as much as possible due to characteristics similar to CMMA circuits.

3.2 Mechanical Features

The complete unit measured 19 x 9.5 x 19 inches consisting of two sub units each mounted in a standard 19 inch rack and each measuring 19 x 9.5 x 9.5 inches.

A bracket on the back of the unit provided in addition to protection for the wire wrap terminals a mounting location for the connectors. The two memory connectors were mounted to this bracket, in addition two test connectors and two I/O connectors were mounted on flexible cables. Each individual board was divided into three sections each section containing the logic required for one CMMA. With this approach the CMMA logic and the wiring list for the multilayer board could be checked prior to final design. Table 3-1 shows the partitioning



TABLE 3-1

AOP BREADBOARD PARTITIONINGPARTITIONED SEGMENTS

BOARD POSITION	CMMA #	X FUNCTION	CMMA #	Y FUNCTION	CMMA #	Z FUNCTION	BOARD TYPE
A	17	ACC(1-3)	17	ACC(4-6)	17	ACC(7-9)	1
B	17	ACC(10-12)	17	ACC(13-15)	17	ACC(16-18)	1
C	17	MQ(1-3)	17	MQ(4-6)	17	MQ(7-9)	1
D	17	MQ(10-12)	17	MQ(13-15)	17	MQ(16-18)	1
E	16	DECODE(MINOR)	14		10		2
F	16	DECODE(MAJOR)	11		12		3
G	4		5		25		4
H	22	SPACK(0-7)	2		3		5
I	22	SPACK(8-15)	6		13		6
J	23		7		1		7
K	15		9		8		8
L	18	ADDER(1-3)	19	MOR(1-3)	20	INT(0-3)	9
M	18	ADDER(4-6)	19	MOR(4-6)	20	INT(4-7)	9
N	18	ADDER(7-9)	19	MOR(7-9)	20	INT(8-11)	9
O	18	ADDER(10-12)	19	MOR(10-12)	20	INT(12-15)	9
P	18	ADDER(13-15)	19	MOR(13-15)	20	REQ(0-3)	9
Q	18	ADDER(16-18)	19	MOR(16-18)	20	REQ(4-7)	9
R	19	ADR(1-3)	21	BLAD(1-4)	20	REQ(8-11)	10
S	19	ADR(4-6)	21	BLAD(5-8)	20	REQ(12-15)	10
T	19	ADR(7-9)	21	BLAD(9-12)	26	LOWER LIMIT Bus(13-14)	11
U	19	ADR(10-12)	21	BLAD(13-16)	26	UPPER LIMIT BUS(15-16)	11
V	19	ADR(13-15)	19	ADR(16-18)	24	CSA,SL	12

TABLE 3-1

AOP BREADBOARD PARTITIONINGPARTITIONED SEGMENTS



of the AOP Breadboard.

3.3 Test Panel and Memory

In order to check out the AOP breadboard a test panel and a 4096 word memory was supplied to Westinghouse by GSFC. To limit interface lines all data transfers between the AOP and the test panel were made through the memory operand register (MOR). Only the MOR and the instruction register were displayed on the front panel. The remaining addressable AOP registers were loaded via the MOR and displayed on the front panel.

Both "read" and "write" operations were permitted with the test panel. Other controls available were: initiating and master clear of AOP, clearing the MOR, clearing and interrupt override control, operation controls. Operational controls were available to initiate one instruction at a time, cycle on one instruction or go into a "run" mode to execute a complete program. Verification was also possible in an "off line" mode of operation.

3.4 Breadboard Checkout

The checkout consisted of exercising the AOP breadboard in an "off line" mode, i.e., without the memory. This procedure checked out approximately 85% of the instructions. After verifying the unit in this manner, the memory unit was interfaced with the AOP breadboard and full instructions were executed and programs were loaded into the memory and executed. The diagnostic programs used to test the AOP breadboard were developed by GSFC. During the initial checkout of the breadboard, personnel associated with the software contractor were on hand to aid in loading and interpretation of the programs. In addition, GSFC personnel assisted in the initial checkout of the breadboard.



4.0 CUSTOM METALLIZED MULTIGATE ARRAYS (CMMA)

An earlier study was conducted jointly by Westinghouse and GSFC to determine availability of Large Scale Integrated Circuits (LSI) from which the AOP could be fabricated. As a result of this study it was decided to take advantage of an existing development program in another NASA facility. The circuits developed under this program consisted of a standard chip containing 92 standard gates, 24 drivers and 24 expanders, 48 resistors and 24 diode clamps. These circuit elements were then interconnected to provide the required logic function. Each different logic function required its own set of custom masks. The CMMA device is discussed in detail below.

The decision to use these circuits was based primarily on the following:

1. The least expensive LSI arrangement available.
2. These circuits allowed maximum usage of circuits on the chip provided the 40 pin limitation of the package was not exceeded.
3. The TTL device is inherently more radiation resistance than certain other circuit types.
4. The speed power product is comparable to or better than other low power circuits.
5. The power dissipation is within the desirable range for space applications.
6. Flexibility enabling the fabrication of a wide variety of logic functions.

4.1 LSI Circuit Description

The CMMA is a two level, bipolar, LSI array utilizing the transistor-transistor logic (TTL) technique. The first layer of metalization interconnects the bipolar devices into logic elements; 92 standard wired-or gates, 24 emitter follower drivers, 24 three input expanders, 48 resistors and 24 diode clamps. The second layer of metalization interconnects the logic elements into functional

CMMA devices.

4.1.1 Die Organization

The LSI Die Organization is illustrated in Figure 4-1. This figure presents standard columns, driver columns, extender columns and bonding pad and scribe areas. The standard columns represent the standard or low power gates illustrated in Figure 4-2. The driver columns represent the driver gates and are illustrated in Figure 4-3. These are the high power driver circuits that are used to interface with circuits on other chips or with other subsystems. Figure 4-4 is the schematic of the circuits in the extender column. Also included with the extender column are resistors and diode clamps. The standard gate circuits are used only inside a given chip, while the driver is used for both internal and external interconnections.

4.1.2 LSI Configuration

The LSI Die Organization illustrated in Figure 4-1 references the basic arrangement of circuits available. The concept utilized on the program for circuit fabrication, is as follows. First a semiconductor wafer, a bipolar array, is processed as a single unit containing many die like those referenced in Figure 4-1. In order that each of these die may become a CMMA device, an interconnect arrangement is required. This leads to the second step. A layer of metallization is deposited on the die which interconnects the bipolar devices into logic elements. A total of 92 standard gates, 24 expanders, 48 resistors, 24 clamps, and 24 drivers are thus formed on each die. These are the building blocks used to functionally implement the circuitry dictated by each drawing. The third step deposits a second layer of metallization to interconnect the logic elements available after the first layer of metallization has been completed. This step creates a Custom Metallized Multigate Function.

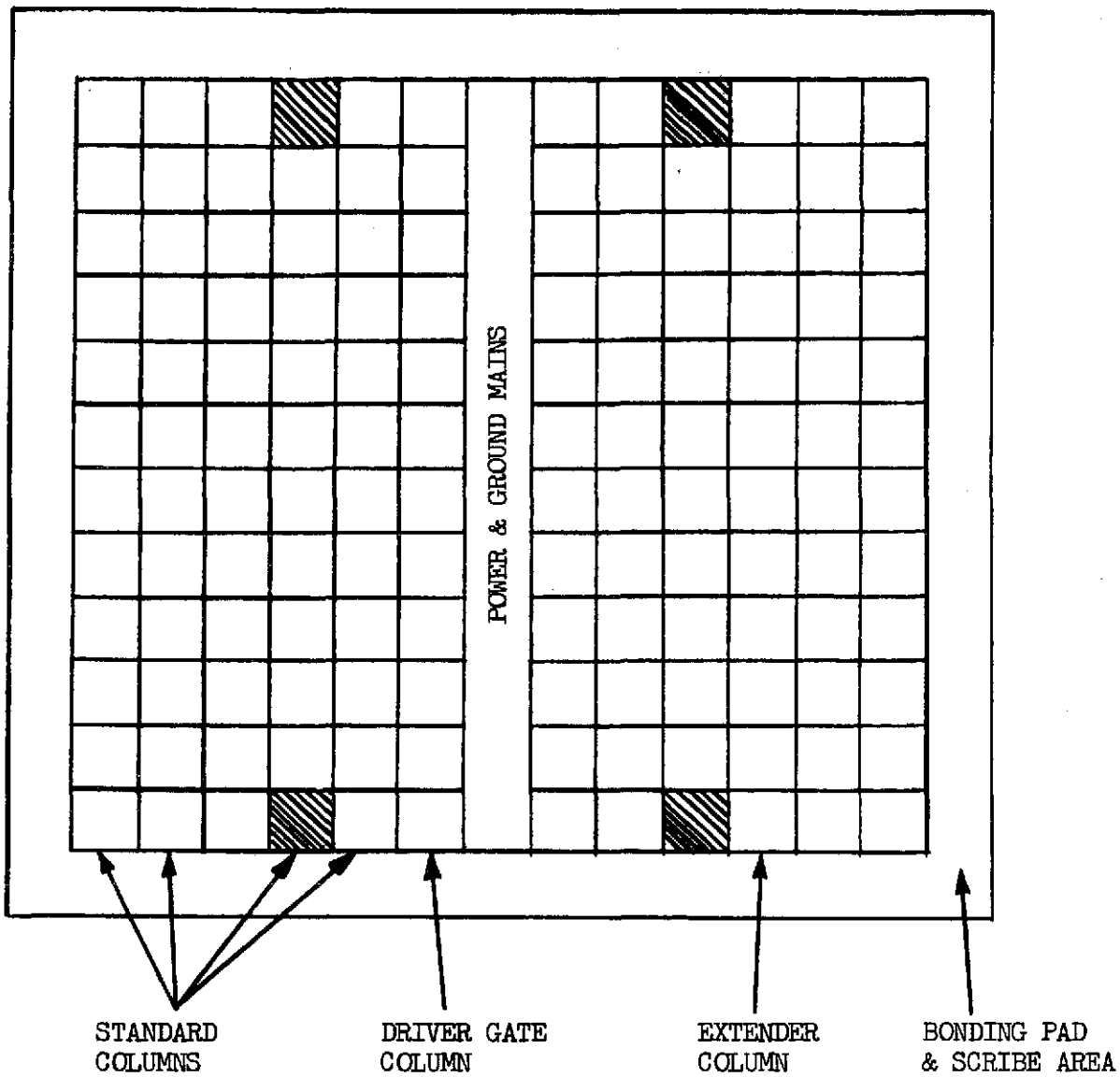


FIGURE 4-1

MA DIE ORGANIZATION

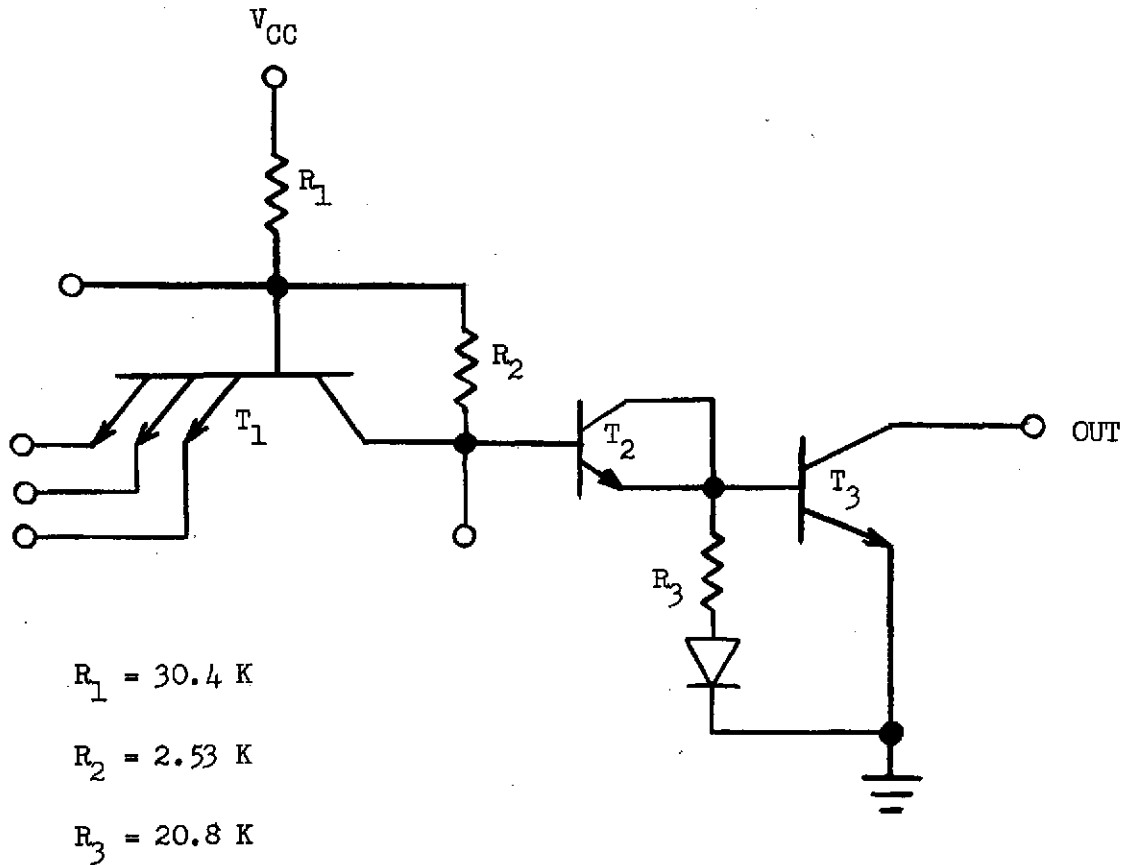
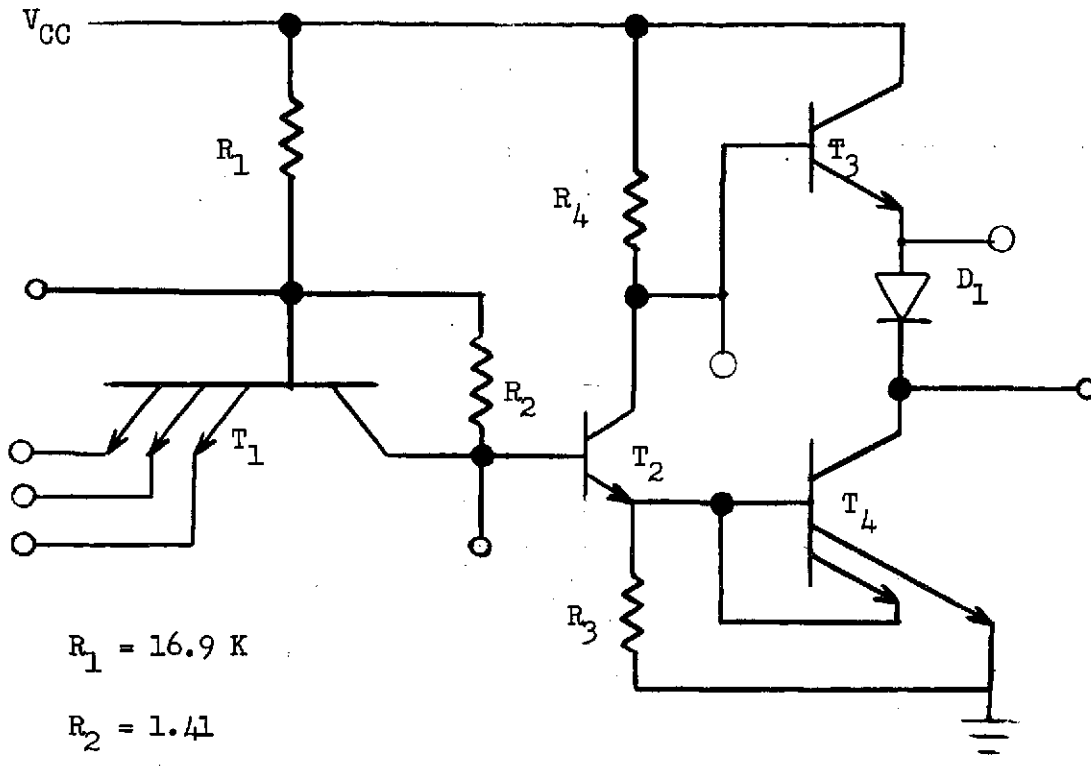


FIGURE 4-2
STANDARD GATE



$R_1 = 16.9 \text{ K}$

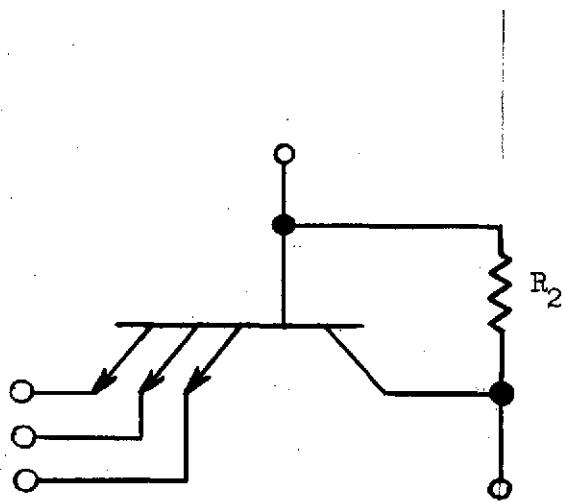
$R_2 = 1.41$

$R_3 = 8.4 \text{ K}$

$R_4 = 10.8 \text{ K}$

FIGURE 4-3

DRIVER GATE



$$R_1 = 30.4 \text{ K}$$

$$R_2 = 2.53 \text{ K}$$

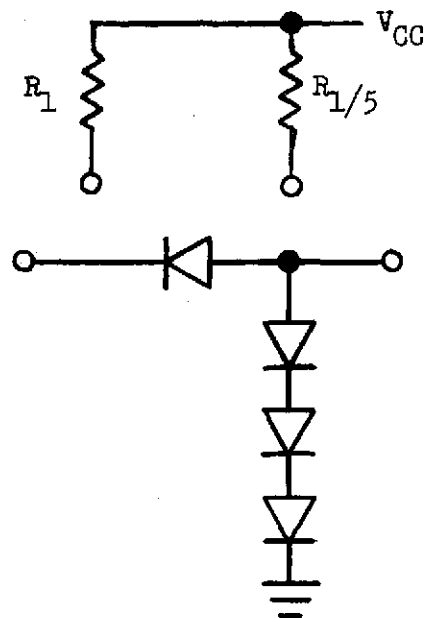
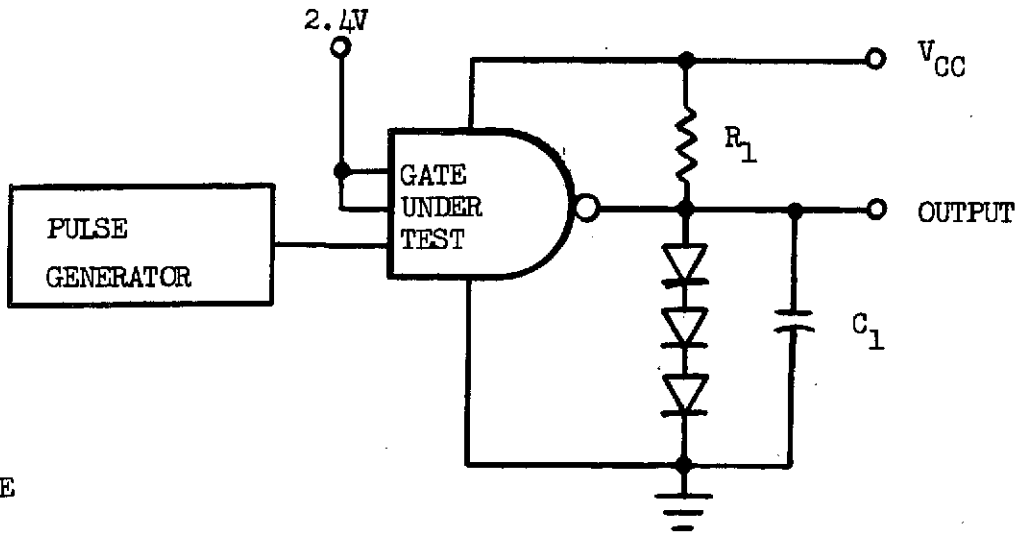


FIGURE 4-4

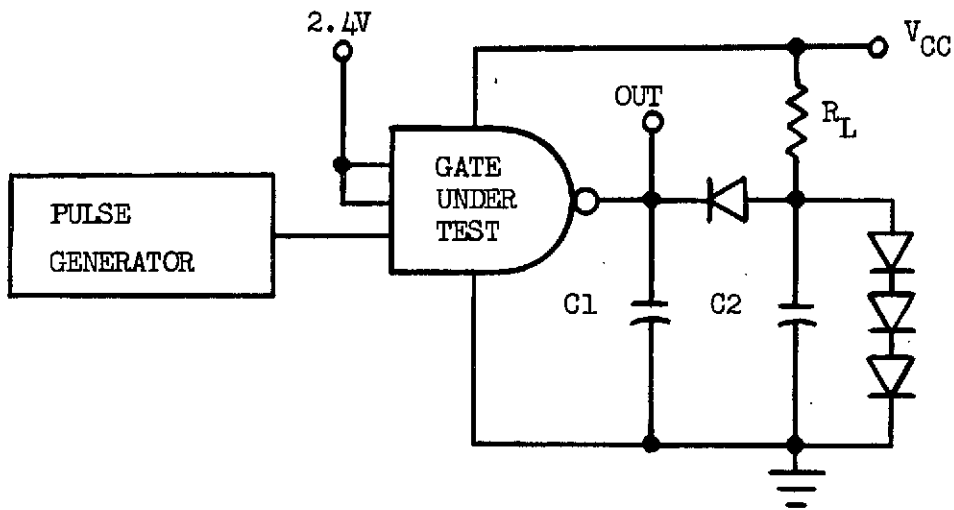
EXTENDER/PULL-UP CELL



a. STANDARD GATE



b. DRIVER GATE



c. VOLTAGE WAVEFORMS

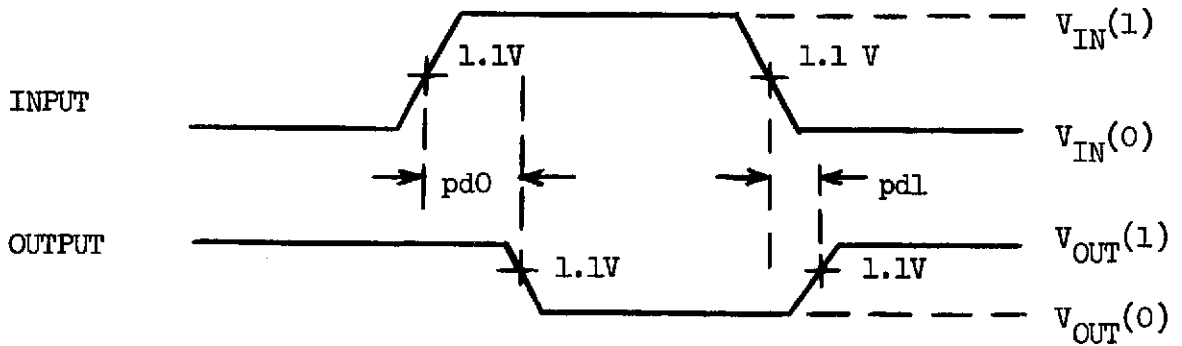


FIGURE 4-5

PROPAGATION DELAY MEASUREMENT SETUP



A via layer or communication layer enabling a larger number of interconnects by passing signals from the first layer to second layer was also associated with the second layer metallization.

4.2 Circuit Layout

The object in designing a system is to partition the logic so that one LSI circuit type can be used for a number of different applications in the system design. In the developed AOP Processor, 27 types and a total of 69 LSI circuits were required.

4.3 CMMA Layout

Westinghouse experience shows that layout problems increase in a nonlinear manner as 100% utilization of logic elements on the die is approached. Also certain circuit configurations requiring complicated interconnects proved difficult.

4.4 Testing

Westinghouse generated two sets of test documents to verify the CMMA circuits. A set of static tests to verify the chip parameters and the logic interconnect. A set of dynamic tests to verify propagation delays and final circuit characterization.

Each chip was probed with forty tips, one for each input/output pad and the static tests cycle was executed. The chip having passed visual prior to static tests was packaged in a forty pin flat pack, given 96 hour burn-in at 125°C. The unit then went through static and dynamic tests before delivery.

4.5 Equation Development and Specifications

In order to design the computer the limitations of the circuits and operating parameter must be well defined. This data was supplied to Westinghouse by the Semiconductor Manufacturer and is shown in tables 4-1, 4-2 and 4-3.



N = Fan-Out (Loading)

M = Additional Outputs Added as Wired-Or Configuration

E = Expanders (Increase Number of Inputs)

R = Additional Pullup Resistors

$$T_{pd1} = 20 + .22 R_L C_1 + \frac{12 M}{N + R}$$

$$R_L = \frac{30}{N + R}$$

$$C_1 = 3 + N + R$$

$$T_{pd0} = 60 + 10 (E - R)$$

$$R \leq E$$

Delay (N = 0-10)

	T_{pd1}	T_{pd0}
a) Fan-Out (N) (Open Collector)	Per T_{pd1} Equation	60 NSEC MAX
b) Wired-Or (Additional Outputs)	Per T_{pd1} Equation	60 NSEC MAX
c) Pullup - No Clamp (Off Chip Drive)	Per T_{pd1} Equation	160 NSEC MAX
d) Expanders	Per T_{pd1} Equation	Per T_{pd0} Equation

TABLE 4-1

STANDARD GATE SPECIFICATIONS (ORIGINAL)



N = Fan Out (Loading)

M = Additional Outputs or Wired-Or Configuration

E = Expanders (Increase Number of Inputs)

R = Additional Pullup Resistor

$$T_{pd1} = 35 + \frac{100}{N(11) + (M+1)(30)} [(5)(M+1) + N + C_1]$$

$$C_1 = 20$$

$$T_{pd0} = 40 + 10(E - R)$$

	Delay (N = 0-50)	
	T_{pd1}	T_{pd0}
a) Fanout	40 NSEC MAX	40 NSEC MAX
b) Wired-Or (Additional Outputs)	Per T_{pd1} Equation	40 NSEC MAX
c) Expanders	40 NSEC MAX	Per T_{pd0} Equation

TABLE 4-2

DRIVER GATE SPECIFICATIONS (ORIGINAL)

STANDARD GATE

Levels Logic "1"	2.4 - 5.0V (Output)	2.0 - 5.5V (Input)	
Levels Logic "0"	0.0 - 0.3V (Output)	0.0 - 0.7V (Input)	
Loading Input Load	1 Unit Load		
Output Drive	10 Unit Loads		
Speed			
T_{pd0}	60 nsec Max (n = 1-10)		
T_{pdl}	Per T_{pdl} Equation		
Current (Input)	.26 ma (-35°C)	.18 ma (25°C)	0.135 ma (100°C)
Power "1"	1.3	1.0	0.77 mw
Power "0"	0.95	0.78	0.71 mw

DRIVER GATE

Levels Logic "1"	2.4 - 5.0V (Output)	2.0 - 5.0V (Input)	
Levels Logic "0"	0.0 - 0.3V (Output)	0.0 - 0.7V (Input)	
Loading Input Load	1.8 Unit Loads		
Output Drive	50 Unit Loads		
Speed			
T_{pd0}	40 nsec Max		
T_{pdl}	40 nsec Max		
Current	.47 ma (-35°C)	.32 ma (25°C)	.25 ma (100°C)
Power "1"	2.5	1.8	1.3 mw
Power "0"	5.2	4.0	3.0 mw

TABLE 4-3

CIRCUIT SPECIFICATION TEMPERATURE RANGE -35°C to +100°C (ORIGINAL)

The ground rule for this data is that a standard gate or driver gate must be clamped with the circuit of figure 4-4 when: four wired "OR" structures are connected to one point or when interfacing with another chip.

These equations and parameters were used in the initial design of the AOP. Subsequently the equations and parameters were modified by the Semiconductor Manufacturer.

4.6 Circuit Characterization

Tables 4-1 and 4-2 refer to equations which define the delays of the standard gate and driver gate. It was very simple to determine the operating conditions of a single gate or driver and thus calculate the specifications but much more difficult to calculate the specifications for a logic chain. It was decided that maximum values could be determined for certain gate combinations. These values were then applied to logic chains which specified a worst case condition.

4.7 Final Circuit Development

In some cases the initial specifications for the circuits and the characterization of the circuits proved to be more stringent than the subcontractor was able to comply with. Changes to the specifications were necessary as the program developed.

4.8 Generation of New Equations

During the program revisions were made to the equations given in tables 4-1 and 4-2 and to the specifications in table 4-3.

In the equation for T_{pd1} , the constant 0.22 was changed to 0.304. When this constant was established the forward drop across the diode V_{BE} was neglected. Assuming the following values,



$$E_{TH} = 1.1 \text{ volts}$$

$$E_O = V_{CC} - V_{BE}$$

$$V_{BE} = 0.8 \text{ volts}$$

$$E = E_O (1 - E^{-t/RC})$$

$$t = 0.304 R_L C_T$$

$$R_L = \text{Total Load} \quad C_T = \text{Total Capacitance}$$

The method of evaluating the capacitance and resistive loads was changed. The capacitive loading on a standard gate output was increased 3 pf for each OFF gate whose outputs were connected to other inputs of the driver gate. Another 1 pf was added for each gate input tied to the OFF gate output. If the OFF gates is the emitter follower type or a resistor pull up is present, no additional capacitance is added.

The value of R is not increased by an OFF gate in the fan-out. The constant 30 in $R_L = \frac{30}{N + R}$ is changed to 39. The final circuit equations are shown in table 4-4 and the final parameters are shown in table 4-5. Devices were delivered to Westinghouse according to these specifications. The power dissipation shown in table 4-5 and in table 4-3 is given for high resistance material. Both low and high resistance material was delivered to Westinghouse. The power dissipation of low resistance material should be increased by 30% over the values given in the tables.

5.0 CMMA BREADBOARD

To facilitate testing the CMMA circuits before mounting them on the multi-layer board, Westinghouse fabricated a CMMA breadboard. It was determined that a breadboard test unit was an essential piece of equipment for testing the chips and evaluating AOP system performance prior to final assembly of the system. The following list of needs and uses for the breadboard supported the decision to build it.

STANDARD GATE

$$T_{pd1} = 20 + .304 R_L C_1 + \frac{12M}{N+R} \quad \text{where } R_L = \frac{39}{N+R}$$

$$T_{pd0} = 60 \text{ nsec } (N = 1-5) \quad C = 3 + N + R$$

$$T_{pd0} = 100 \text{ nsec } (N = 6 - 10)$$

$$T_{pd0} = 60 + 10 (E - R) \quad (N = 1-5)$$

Where Applicable

$$T_{pd0} = 100 + 10 (E - R) \quad (N = 6-10)$$

DRIVER GATE

$$T_{pd1} = 50 \text{ nsec } (N = 1-50)$$

$$T_{pd0} = 50 \text{ nsec } (N = 1-50)$$

$$T_{pd1} = 35 + \frac{100}{N(11) + (M+1)(30)} (5(M+1) + N + C_1)$$

Where Applicable

$$T_{pd0} = 50 + 10 (E - R)$$

TABLE 4-4

REVISED EQUATIONS

STANDARD GATE

Levels Logic "1"	2.4 - 5.0V (Output)	2.0 - 5.5V (Input)	
Levels Logic "0"	0.0 - 0.5V (Output)	0.0 - 0.7V (Input)	
Loading Input Load	1 Unit Load		
Output Drive	10 Unit Loads		
Speed			
T_{pd0}	60 nsec Max	(n = 1-5)	
	100 nsec Max	(n = 6-10)	
T_{pd1}	Per T_{pd1} Equation (Table 6)		
Current	2.6 ma (-35°C)	.18 ma (25°C)	0.135 ma (100°C)
Power "1"	1.3 mw	1.0 mw	0.77 mw
Power "0"	0.95 mw	0.78 mw	0.71 mw

DRIVER GATE

Levels Logic "1"	2.4 - 5.0V (Output)	2.0 - 5.0V (Input)	
Levels Logic "0"	0.0 - 0.5V (Output)	0.0 - 0.7V (Input)	
Loading Input Load	2.0 Unit Loads		
Output Drive	50 Unit Loads		
Speed			
T_{pd0}	50 nsec Max		
T_{pd1}	50 nsec Max		
Current	.47 ma (35°C)	0.38 ma (25°C)	0.25 ma (100°C)
Power "1"	2.5 mw	1.8 mw	1.3 mw
Power "0"	5.2 mw	4.0 mw	3.0 mw

TABLE 4-5

CIRCUIT SPECIFICATION TEMPERATURE RANGE $\approx 35^{\circ}\text{C}$ to 100°C (FINAL)



1. The program schedule called for the delivery of the last CMMA chips before final delivery of the multilayer board. The CMMA breadboard allowed system testing without the PC board and consequently made it possible to deliver the system earlier.
2. Troubleshooting was much easier when questionable chips could be changed without soldering or cutting leads.
3. Any changes required to the multilayer board could be made before the chips were mounted, thus avoiding exposure of the circuits to excessive handling.
4. Verification of the wiring list used to layout the multilayer board, therefore, all corrections to the board could be made prior to the assembly of the engineering model.
5. The interfaces between the AOP-CPU, plated wire memory, and special I/O could be tested and debugged prior to final assembly of the system.
6. The circuits could be tested for both correct functional operation and speed.
7. Any new system problems arising from the use of CMMA devices could be determined and necessary design changes could be easily evaluated.

In practice the breadboard proved to be an invaluable piece of equipment without which it would have been extremely difficult if not impossible to debug the system in the time allocated.

5.1 Fabrication

The fabrication technique of the CMMA breadboard was influenced by the availability of parts and time required for assembly which led to the selection of a dual bucket assembly with a wire wrap back panel similar to the original breadboard. Westinghouse designed a double sided printed circuit board which



plugged into the buckets and on which contactors were mounted. These contactors were used to hold the CMMA circuits.

5.1.1 Contactors

The contactors for the CMMA breadboard were selected after observing the test set used at the subcontractor's facility. The CMMA circuits were packaged in the 50 mil lead packages and then placed in a 50 pin plastic carrier by the subcontractor. The circuits were left in the carriers for test and shipment to Westinghouse. The subcontractor also utilized a 50 pin contactor to firmly hold the 50 pin plastic carriers and at the same time make electrical contact with the leads of the CMMA package. The contactor contained 50 pins which extended from the bottom and were used to make electrical connections with the outside world. A solid metal latch pivoted on one end of the top of the contactor and was used to secure the plastic carrier and in turn force the package leads onto the contactor pins without damaging the leads of the package.

Utilizing these contactors, Westinghouse was able to insert the circuit, as delivered by the subcontractor into the system for test purposes. Any package could be placed in any contactor, facilitating the system checkout.

5.1.2 Printed Circuit Board

After selecting the contactor for the CMMA breadboard a board and bucket arrangement similar to the original breadboard was used to efficiently package the CMMA breadboard. The boards used for this purpose were designed by Westinghouse and were capable of holding three contactors and 120 output pins making them identical at the interface with the original breadboard. The board edge was plated and utilized as an edge connector to mate with the back panel mother plate. Since 66 circuits were utilized in the original system, and eventually increased to 69, 23 boards were required to implement the design. As the design and test progressed, an additional board was added, for

the additional type 27 circuit.

5.1.3 Board Rack and Back Panel

Two racks, each capable of holding thirteen boards (identical to the racks used for the first breadboard) were used to mount the printed circuit boards and provide a wire wrap back panel for interconnecting the boards. Twenty-three of the twenty-six available 120 pin connectors were used for interconnecting the CMMA circuits. All wires to the test panel, memory, or special I/O were routed to the remaining three connectors to provide a localized area for interface signals. The connectors for interface cables were mounted on a panel at the back of the unit as in the original breadboard and the back panel was automatically wire wrapped.

5.1.4 Wiring List

The goal in the generation of the wire list was to generate one list for both CMMA breadboard and the engineering Model AOP. The wiring list for the back panel was designed so that each 120 pin connector was sub-divided into three forty pin sections. Within the forty pin sections the pin numbers were assigned identical to the CMMA package pin numbers. This made it possible to generate a breadboard wire list similar to the engineering model wiring list automatically with a special computer program. This feature also facilitated the breadboard checkout. The leading item in the wire list generation was the "pin out" information generated by the subcontractor. Westinghouse had the wire list program available long before pin out data was available.

5.2 Testing

When the CMMA breadboard had successfully completed all the tests that could be practically performed at Westinghouse, it was taken to NASA GSFC for further testing to verify the AOP plated wire memory, the diagnostic tests and the AOP I/O. The advantages of performing these tests while the AOP was still



in breadboard form include the following:

1. The residual spare chips to be delivered to GSFC at the termination of the contract could be conveniently tested.
2. The speed of the system could be measured, critical paths determined, and any design changes to improve performance could be easily implemented and evaluated.
3. GSFC system integration between the CPU, Memory, I/O and software could be performed at an earlier date than would otherwise have been possible.

The tests performed at GSFC with the aid of the computer system and I/O were more exhaustive than those conducted at Westinghouse. It was practical to load much longer and more comprehensive diagnostic programs through the I/O, under computer control. Memory contents could be examined and compared with the program originally loaded to diagnose failure quickly. Interrupts, I/O cycle steal operations and CPU programs were run concurrently over a wide range of I/O rates to thoroughly exercise the memory control hardware.

6.0 ENGINEERING MODEL AOP

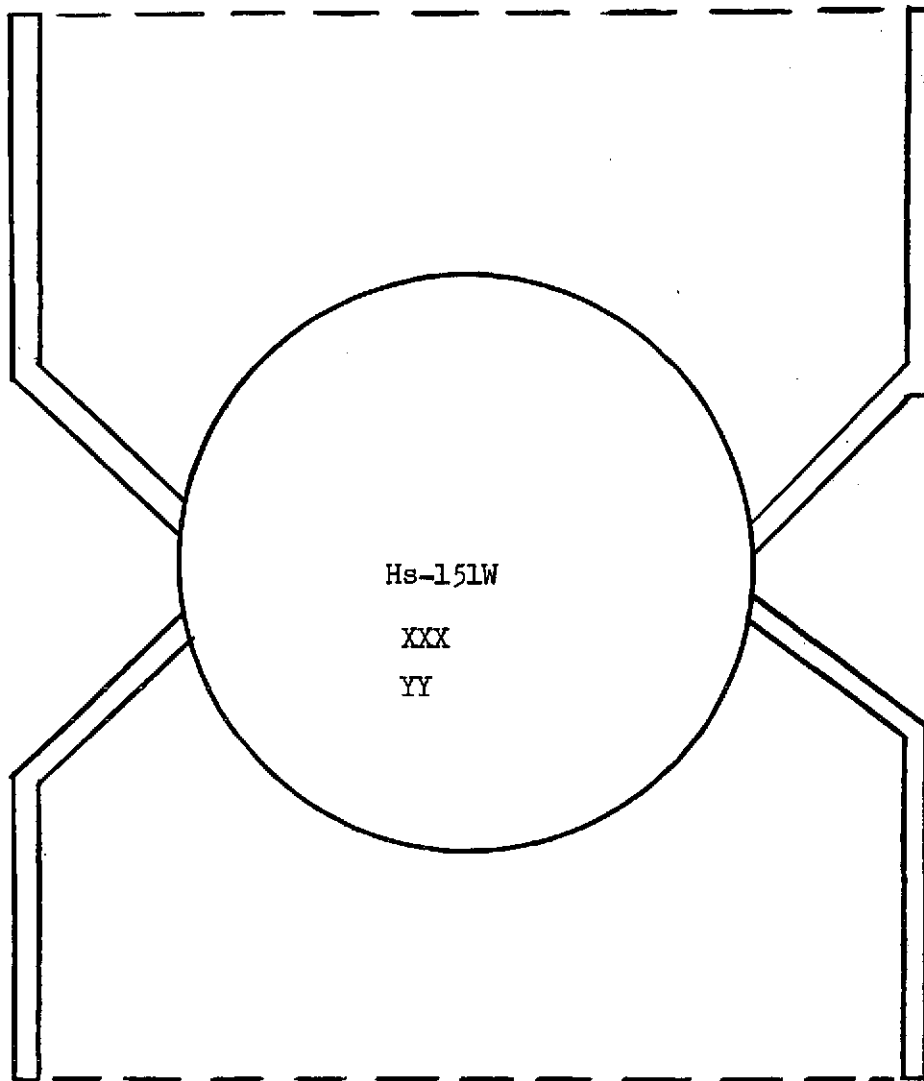
The engineering model AOP was designed and fabricated to resemble, as closely as possible, a flight quality processor. The AOP was designed utilizing the latest multilayer board technology, an ISI circuit technology and a compressed packaging arrangement. The AOP consists of a chassis with a volume of 112 cubic inches, containing four cutouts for four 100 pin Deutsch connectors, mounting positions for the AOP multilayer P.C. Board and the special I/O Board. The design also included an eighteen layer multilayer board, 69 CMM circuits, a special I/O mating connector and two capacitors (one to stretch MINIT -320 pf and one to delay the IRSCLK -680 pf). The chassis is referenced by drawing IRD 2857 and the board layout by drawings RDDOOL-A and RDDOOL-J.



6.1 CMMA Package

Section 4 discusses the CMMA circuit technology and development concept. The basic CMMA circuit consists of a bare chip approximately 150 mil x 175 mil which, in the AOP program, was to be packaged in one of two 40 pin packages. The basic package configuration was that of a circular flat pack approximately $\frac{1}{2}$ inch in diameter with twenty leads extending from opposite sides. Figure 6-1 illustrates the package and associated markings. The packages used in the system were supplied by Sprague and the larger and smaller units had part numbers FH115 and FH80 respectively. The difference in the two packages were in the lead dimensions only; the FH80 package was organized with 35 mil center to center spacing in the leads. The overall package size is approximately 1.0 inches by .7 inches. The larger package (FH115) was designed with 50 mil center to center lead spacing. The overall package size is approximately 1.4 inches by 1.0 inches.

Originally, the FH80 package appeared to be the most desirable from a size and therefore packaging standpoint, the overall dimensions of this package would allow for a more dense packaging approach and a smaller multilayer board. While these packages appeared desirable from a packaging standpoint, other factors indicated the larger package was more desirable for a development type program. First, the subcontractor was utilizing the FH115 package in the pre-development phase and these packages were available at his facility. Secondly, the subcontractor's test jigs and other fabrication setups were organized around the larger package and therefore it would be more expensive to use the smaller package. Thirdly, many of the layout facilities available for multilayer board layouts were organized on divisions of 100 mil center, i.e. 50 mil centers were quite desirable. Finally, both the ease in the initial soldering and the capability of removing CMMA circuits from the solder



XXX: DATE CODE

YY: CIRCUIT NUMBER

QUANTITY: _____

FIGURE 6-1

CMMA PACKAGE

pads dictated the use of the larger package. The final system required 69 of the CMMA packages to implement the AOP design.

6.2 Multilayer Board Fabrication

The design for the AOP included the latest multilayer board (MLB) technique, with a design goal to mount the CMMA circuits on one multilayer board, utilizing both sides of the MLB. The acceptable maximum was to use two (2) MLB's in the construction of the AOP in order to package the CMMA circuits.

6.2.1 Multilayer Board Types

Two available techniques for multilayer board construction were considered acceptable, the post build up technique and the plated-through-hole technique. Both techniques require artwork to define the printed circuit wiring for the various layers and the hole layout.

The post build up technique was being evaluated by GSFC under a separate contract for the AOP as well as other programs. The company supplying GSFC with these boards discontinued their efforts in this area; therefore, Westinghouse chose the plated-through-hole technique for the AOP multilayer board. The plated-through-hole technique had been widely used in industry and was considered a reliable packaging and wiring technique.

The fact that each CMMA package contained 40 pins required that 40 holes must be drilled through the board, and with CMMA's mounted on both sides, 80 holes per square inch were necessary. This density of drilled holes is considered to be the main disadvantage of the plated-through-hole type multilayer board. The hole area and associated clearances consumed much of the printed circuit area, thus requiring additional layers to complete the assembly.

6.2.2 Multilayer Board Configuration

The multilayer board configuration was controlled by two factors, the



available area for the board and the number of circuits to be packaged. As previously stated, it was possible to use two multilayer boards for the AOP; but this was not the design goal. After several layouts and consultations with GSFC, the final board configuration for the system was selected. One board contained 35 circuit positions on either side in a rectangular 5 x 7 array. One edge of the board contained the lower and upper bus connector areas while the other edge of the board contained the test panel and I/O connector areas, the short dimension of the board contained a connector area for the special I/O board in case it should be mounted inside the AOP chassis. This board size dictated that the chassis size be increased to 8.95 inches by 6.9 inches, exceeding the outline of the memory unit.

The packages were evaluated and it was decided a 1" x 1" area would be sufficient to package one CMMA circuit, requiring the leads to be cut to approximately .25 inches max. in the center.

The design and assembly of the multilayer board included a connector area in which a 70 pin connector was mounted. This connector was to be used at some later date to test out various special I/O designs that might be required for system applications. The basic board layout is illustrated in the layout drawings for the AOP, RDDOOL-A and RDDOOL-J. All chip types and the associated AOP functions are included. Connector hole areas are also shown on these drawings. The basic board layout became a 5" x 7" area with circuits on both sides and the remaining area was to be used for connectors, holes and individual components. One spare position E2 exists on the multilayer board with no connection to any of the forty pins. Position E3 was originally a spare position but an additional type #27 circuit was included in the design when incorporating changes to improve system speed.



6.2.3 Multilayer Board Artwork

The fabrication of the multilayer board (MLB) was initiated with the generation of artwork which defined the hole pattern and printed circuit layout for the MLB. This artwork was used as a mask in the etching process which defined the various layers of the MLB.

The artwork for the AOP MLB was generated by computer from a program designed for this purpose. Westinghouse generated a wire list which totally defined the interconnect structure between the CMM pins and the connector areas defined on the MLB layout drawings.

The original goal was to limit the number of layers to a total of 14, including a power and ground plane. An initial evaluation by the company generating the artwork indicated that at least sixteen layers would be required to assemble the MLB. The final configuration required 18 layers and the following set of ground rules:

1. The computer program must be given complete freedom in the positioning of the circuits on the board.
2. The computer program must have complete freedom in assigning signal names to the hole positions within each connector area.
3. The 5 x 7 circuit array would be shifted in the long dimensions so that there would be an area for holes between every two rows of circuits if necessary.
4. The top and bottom layers would be used for printed circuit wiring as required by the computer program.

6.3 AOP Chassis

The AOP chassis was to be identical, in perimeter, including mounting holes, to the plated wire memory developed for the AOP system. The original outline was a rectangular configuration with dimensions of 8.95 inches by



5.85 inches. In each corner of the unit were two holes, one for a retaining rod and one utilized to mount a similar unit on top. The basic premise behind the AOP system design was to "stack" memories and processors to form a finished system.

With the design of the AOP reaching a final stage, it was known that 66 to 70 circuits would be required for the AOP design. In order to package these circuits within the confines of the dimensions specified by the plated wire memory, five separate layouts were undertaken for the multilayer board. The net result of this effort was an undesirable packaging arrangement, finally, the dimensions for the multilayer board were expanded to approximately 7.9 x 5.9 inches.

The initial design of the chassis included variations of reversing the mounting flange on one side and then both sides of the chassis. After a number of different configurations were generated, it was decided to increase the overall dimension of the chassis with the form factor similar to the plated wire memory chassis. This decision resulted in the AOP chassis increasing to 8.95 inches by 6.90 inches. With the 8.95 inch dimensions being maintained, the plated wire memory could be mounted on the AOP for system tests. Any additional plated wire memories were to have their form factor modified to coincide with the dimensions of the AOP. The two long sides of the AOP unit were utilized to mount four -100 pin Deutsch connectors. The "lower bus" connector and the "test panel" connector were assembled with the unit. The position for the "upper bus" connector was left vacant so that GSFC could mount a 100 pin connector that would mate with the special I/O board as a lower bus connector. The additional connector position was for a future I/O interface connector.

The AOP chassis is approximately 1 3/4 inches in height, designed so that



two multilayer boards could be mounted in the unit. The board positions were located on the top and bottom area of the chassis with the center area between the boards allocated for the connectors and the associated wiring. With the Westinghouse effort resulting in only one multilayer board required for the system design, the additional position could be utilized by GSFC for their special I/O boards. To facilitate this arrangement, Westinghouse included the special I/O connector area on the MLB when the artwork was being generated.

It is feasible to reduce the overall height and thus the volume of the AOP by either selecting different connectors, and/or extending the connectors beyond the board area. In both cases the height of the chassis could be reduced since the total height required for the two boards is much less than the 1 3/4 inches allocated for the chassis height. Drawing IRD 2857 references the AOP chassis.

6.4 Assembly

After checkout of the CMMMA breadboard at GSFC the unit was returned to Westinghouse and the CMMMA circuits were removed and readied for mounting.

6.4.1 Multilayer Board Changes

The tests conducted at GSFC on the AOP resulted in the addition of a second type 27 circuit to the final design as an aid in achieving the desired results.

The speed evaluation to determine methods of increasing speed resulted in the following changes to the multilayer board. These changes consisted of cutting the printed circuit wiring on the board to free the circuit lead pad from the remaining multilayer board wiring. The circuits were then soldered to the board and wires were soldered to the top of circuit leads. These wires were then attached to the desired leads on other circuits. If plated-through-holes were available they were utilized to attach the wires to the board.



Table 6-1 illustrates the changes incorporated on the multilayer board to improve the system speed.

6.4.2 CMMA Circuit Assembly

The CMMA circuits were mounted on the multilayer board by parallel gap soldering techniques which were currently in use on another space program at Westinghouse.

The leads are first bent in a 1" bending tool, cut to the proper length, tinned and soldered to the multilayer board with a parallel gap weld/solder machine.

Following the circuit mounting, all standoffs, connectors and connector wiring were mounted on the multilayer board. In order to reduce possible noise on the voltage bus, four bypass capacitors were mounted on the board, two on the unused upper bus connector area and two in the area below the I/O connector opening in the chassis.

Since the plated-through-holes were located directly under each CMMA package, a mylar insulating strip was cut and was adhered to the bottom of each circuit. This insulating layer completely isolated the package from the holes. This technique is common practice on several Westinghouse space programs.

6.5 Tests

After the initial tests were conducted on the CMMA circuits the AOP was completely assembled and "OFF LINE" tests were conducted on the unit with the results being very favorable.

All initial tests were conducted on the AOP unit with the diode clamp and 10K resistor pullup connected to the set lines associated with the MOR set inputs. The tests conducted were successfully executed within a seven micro-second "ADD" time, in fact, the unit performed the diagnostic tests supplied



<u>REFERENCE SIGNAL</u>	<u>SIGNAL NAME</u>	<u>MODIFICATION</u>
IRCLK	IRCLK	Remove pad from D06/30
	IRCLK	Add wire from E06/11 to D06/15
	DISPLD-	Add wire from D06/35 to D06/16
	DISPLD-	Add wire from D06/16 to D06/17
	IRSCLOCK	Add wire from D06/19 to J06/24
AIRCLK	Ø SYNC	Remove pad from D06/10
	Ø PROC	Add wire from I06/22 to D06/10
PAGE	Ø SYNC	Remove pad from C07/9
	Ø SYNC	Remove pad from G07/9
	Ø PROC	Add wire from D06/33 to C07/9
	Ø PROC	Add wire from C07/9 to G07/9
PAGE ADR	ADRCLK	Add wire from A04/30 to A05/30
	ADRCLK-	Add wire from A04/30 to E03/30
	ADRCLK-	Add wire from E03/29 to E01/6
MORCLK	MORCLK	Add wire from F06/4 to E03/3
	MORCLOCK	Add wire from E03/4 to A03/6
OUTTO	OUTTO-	Add wire from E04/12 to E03/31
	OUTTO	Add wire from E03/32 to E07/6
#27	GND	Add wire from E03/20 to GND (E03/20)
	5 volts	Add wire from E03/40 to K00/11

TABLE 6-1

MULTILAYER BOARD WIRING CHANGES



by GSFC with an ADD time approximately 5.5 microseconds. The major difficulties encountered were with the limits specified for the "readcomplete" signal, 350 \pm 100 microsecond. This dictated the machine must operate successfully over the temperature range with a 250 microsecond readcomplete pulse and tests conducted indicated this was not feasible mode of operation.

The unit properly functioned over the temperature range with the 500 microsecond "readcomplete" pulse generated by the GFE memory with the pullup removed. The ADD time did not exceed 6 microseconds over the full temperature range. Since the arrival of "readcomplete" had been delayed to 750 \pm 50 microseconds no problems were encountered in this area.



APPENDIX

ADVANCED ONBOARD PROCESSOR (AOP)

Description and Instruction Set



AOP SYSTEM DESIGN

1.0 INTRODUCTION

The Advanced Onboard Processor (AOP) is a small, lightweight, low power, stored program digital computer intended for onboard spacecraft applications. It achieves its power and size objectives through the use of low power large scale integrated (LSI) circuit technology, multilayer board techniques, and a low power plated wire memory. The AOP evolved from an earlier model, the Onboard Processor (OBP), which was developed jointly by Westinghouse and NASA/GSFC as a general purpose processor on earth orbiting satellites. The development of the AOP resulted in a reduction in size, weight, and power consumption compared to the OBP. Figure A-1 illustrates the AOP unit.

The AOP processor module contains both a powerful CPU and a flexible I/O unit requiring only a memory unit and power supply unit to form a complete general purpose computer system. Figure A-2 represents the stacked arrangement representing the final system configuration. The CPU, the control for interrupts and I/O direct memory access (DMA) channels and the memory bus controller are packaged as one unit (one board). The remainder of the I/O, being that section of the I/O which is application dependent, is packaged separately (one board) but within the same physical module. Figure A-3 is a representative block diagram of the AOP - CPU. I/O units in general are required to change from mission to mission to provide different functions, interfaces, data rates, etc. However, much of the I/O system, namely that part which interfaces directly with the CPU and memory, tends to remain unchanged no matter how the application requirements vary. Greater flexibility was achieved and a wider variety of applications were served when the section of the I/O which must be continually modified was minimized and segregated

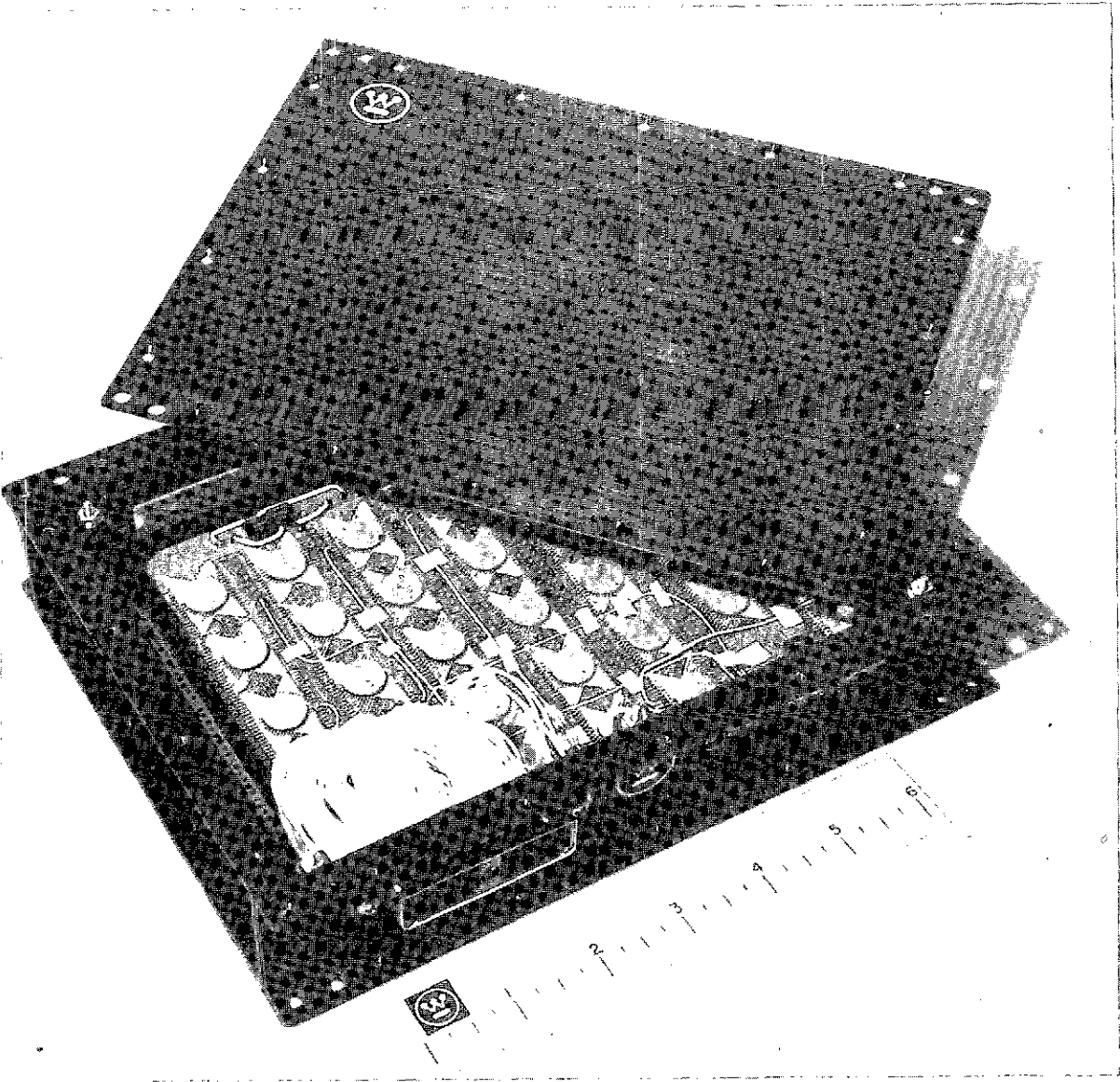


FIGURE A-1 AOP UNIT

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A-3

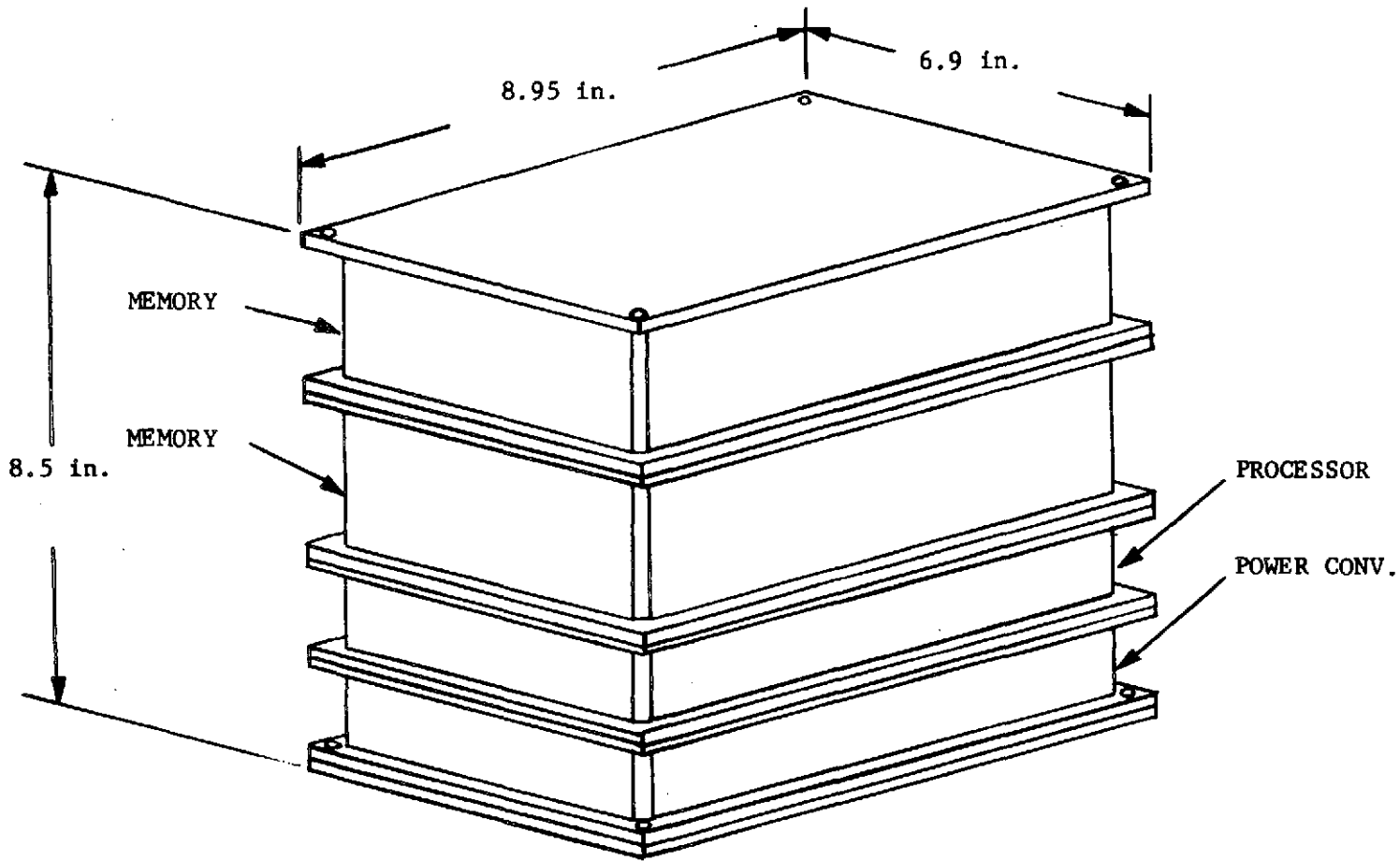


FIGURE A-2 TYPICAL AOP STACKED CONFIGURATION



Aerospace Division

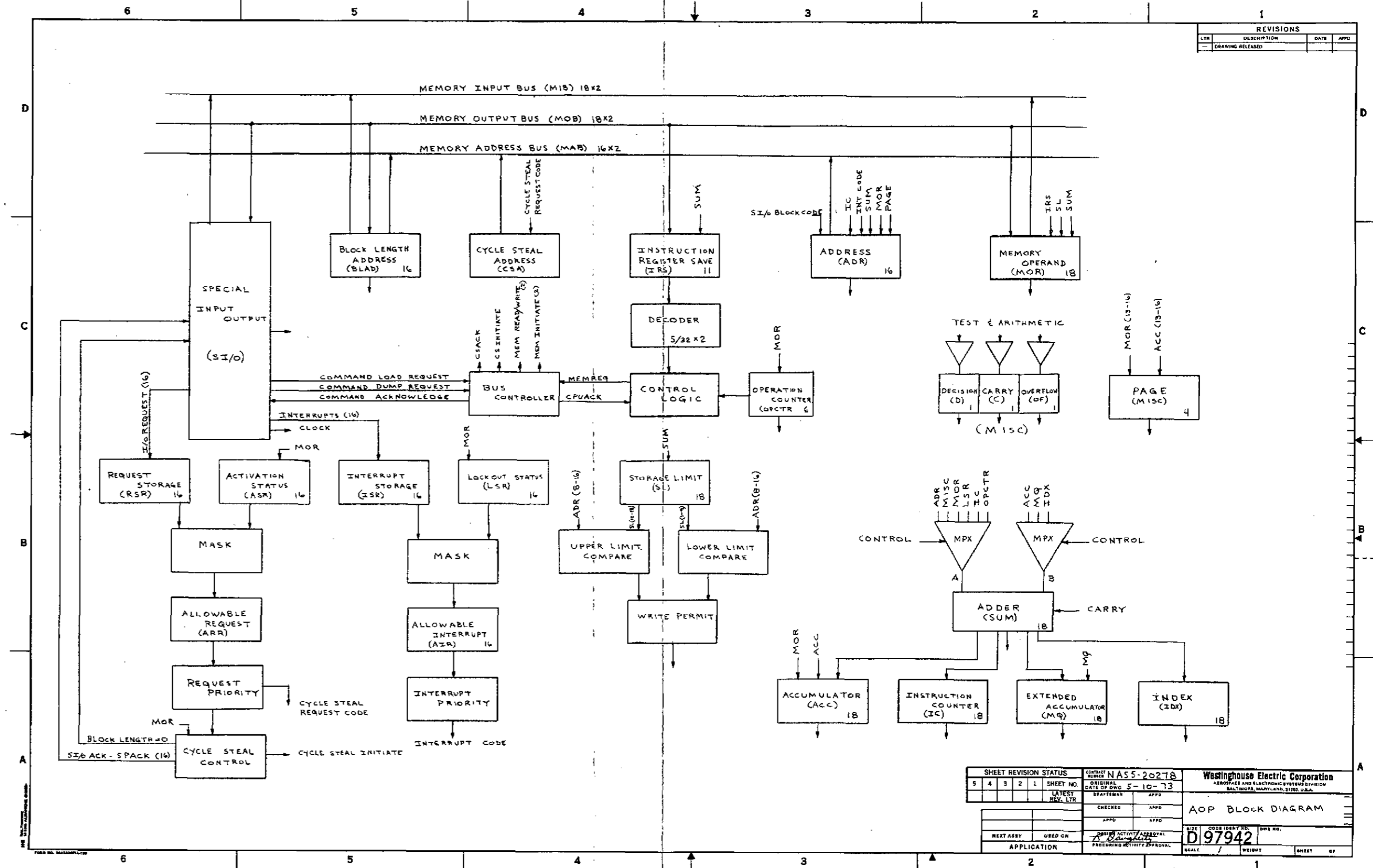


FIGURE A-3 AOP BLOCK DIAGRAM

R

B

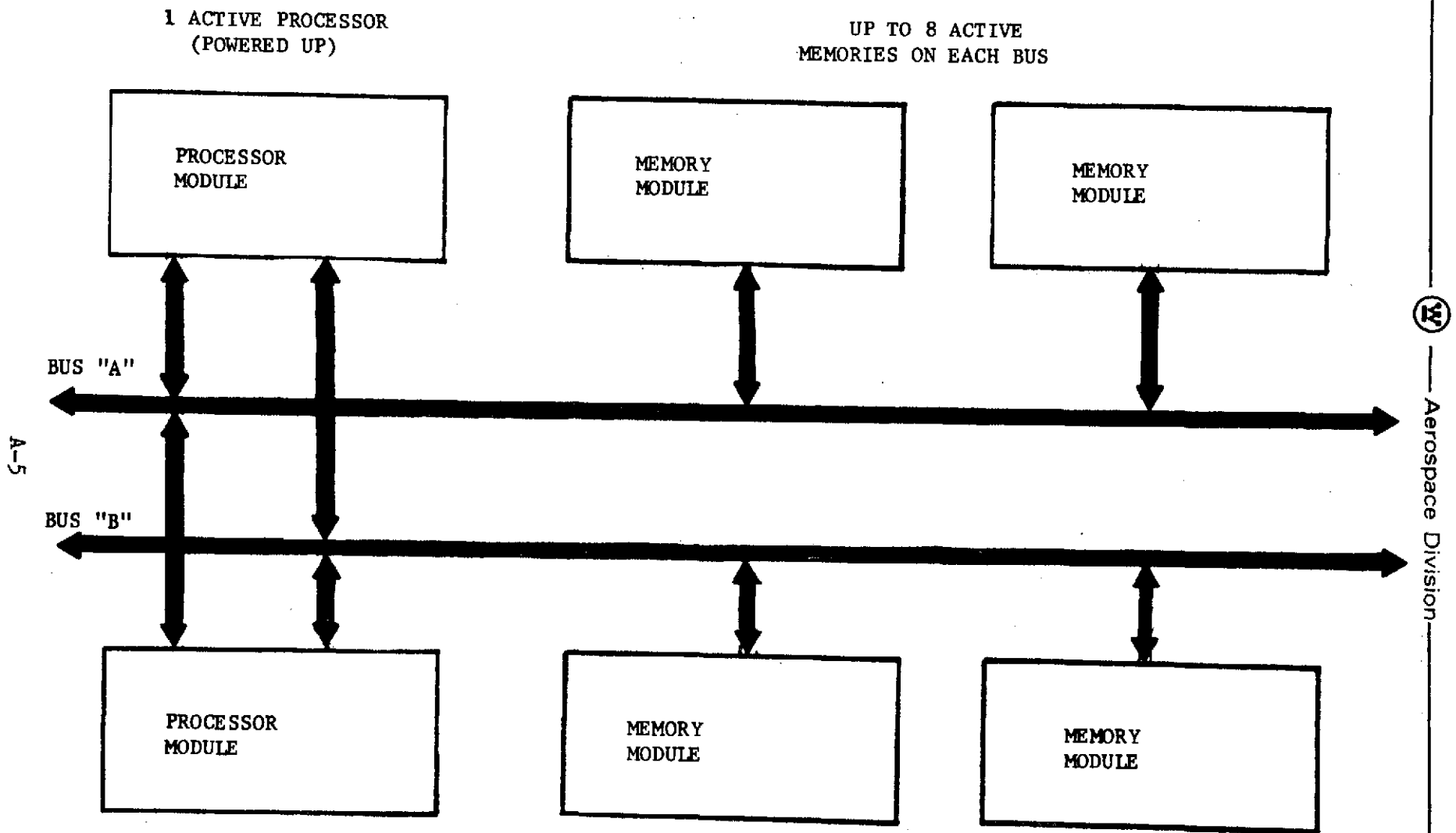


FIGURE A-4 AOP SYSTEM CONFIGURATION



1. WORD LENGTH - 18 Bits
2. INSTRUCTION WORD - 18 Bits
3. DATA FLOW - Parallel
4. DATA TYPE - Fixed Point - Fractional - 2's Complement
5. NO. OF INSTRUCTIONS - 55
6. ADD TIME - 5.0 Microseconds
7. MULTIPLY TIME - 38 Microseconds
8. DIVIDE TIME - 75 Microseconds
9. CLOCK RATE - 800 KHz
10. NO. OF INDEX REGISTERS - One
11. OPERATIONS PER SECOND - 200K
12. INDIRECT ADDRESSING - Load and Store
13. DOUBLE LENGTH ACCUMULATOR
14. TECHNOLOGY - TTL-LSI
15. SIZE - 112 cu. in. (including all I/O)
16. WEIGHT - 3 pounds (including all I/O)
17. POWER CONSUMPTION - 5 Watts (CPU-Fixed I/O)
18. INTERRUPTS - 16 multilevel interrupts with program lockout control and interrupt override feature.
19. DIRECT MEMORY ACCESS (DMA) - Up to 16 devices time share one channel maximum I/O rate - 66K words per second.
20. COMMAND LOAD AND DUMP-GROUND COMMAND - Control for direct memory load or dump maximum load word rate - 300K words per second.
21. MEMORY CAPABILITY - 4K modules expandable to 64K (each module incorporates cycle by cycle power switching).
22. Direct Addressing of 4096 words in any page (addressing of all 64K of memory is accomplished by indexing or paging).
23. MEMORY WRITE PROTECTION - Storage area assigned in 128 word blocks.
24. Busing Organization for parallel unpowered spare processor and/or memory modules.

TABLE A-1

AOP CHARACTERISTICS



from the rest of the system. At the same time, the I/O DMA channels were re-designed to permit a reduction in the hardware required to implement them.

Table A-1 specifies the AOP characteristics.

To meet the stringent reliability requirements of space hardware the AOP was designed utilizing a dual bus concept. In operation, at least two 4096 word memory modules would be used, one connected to each bus. All data, address and control lines for the memories are duplicated so that in the event of a failure in one bus, the remaining memory can be programmed and the mission can continue although possibly in a degraded mode. As many as eight memory modules may be connected to each bus. In addition, an unpowered standby processor module may be connected to both buses without affecting the operations of the powered processor. Figure A-4 illustrates an AOP system busing organization.

The circuit technology selected for the AOP was that of a custom metalized multigate array (CMMA) LSI circuits containing an array of 116 three input NAND gates and 24 three-input expanders which may basically be connected in any manner. The gates are divided into two groups; 92 low power open collector standard gates which are used to drive inputs on the same chip, and 24 high power active pullup driver gates used to drive inputs on other chips or outside the processor module. The driver gates may be connected in such a way as to convert them to open collector output devices for such uses as memory bus drivers where the "wired-or" function is necessary. The processor is partitioned into 69 CMMA circuits of 27 unique types. The circuits are mounted in circular 40-pin flat packages which are mounted on both sides of a single 18-layer multilayer board. The board is mounted in an aluminum chassis 8.95 x 6.90 x 1.76 with space allocated for one other board containing the SIO circuitry.



2.0 CPU OPERATION

The AOP CPU is a fractional, two's complement fixed point computer. The word length is 18 bits. All logical and arithmetic operations and register transfers are fully parallel. There are four directly addressable registers and two additional registers which can be addressed via special instructions.

2.1 REGISTERS

An 18 bit accumulator (ACC) is used as one operand in most arithmetic and logical instructions and also stores the result of these operations. It's contents may also be tested for odd parity, compared to memory contents, complemented, negated, tested for sign and reversed.

The 18-bit memory operand register (MOR) is a storage register for all data being transferred to and from memory. It also serves as the other operand in arithmetic and logical operations and acts as temporary storage during register exchanges.

An 18-bit address register (ADR) stores the effective address for all memory accesses in its least significant 16-bits. It may be loaded from the MOR, the adder, and the page register. Also a four-bit block code specifying a particular 4096 word block of memory may be loaded into the ADR from the SIO during an Interrupt, and EXIT or RESUME instruction.

An 18-bit extended accumulator or multiplier-quotient register (MQ) may be loaded from and stored into memory. It holds the least significant half of the divisor and the entire 18-bit quotient during a DIVIDE instruction. It may be shifted with the accumulator during DOUBLE SHIFT, DOUBLE CYCLE and NORMALIZED instructions. It retains the least significant half of the product in the MULTIPLY instruction.

An 18-bit index register (IDX) may be added to the ADR to form the effective



address for an indexed memory operand instruction. It may also be loaded from and stored into memory.

An 18-bit instruction counter (IC) stores the effective address for the next instruction fetch. It is updated during every instruction fetch and loaded into the ADR at the end of each instruction (with the exception of branching instructions).

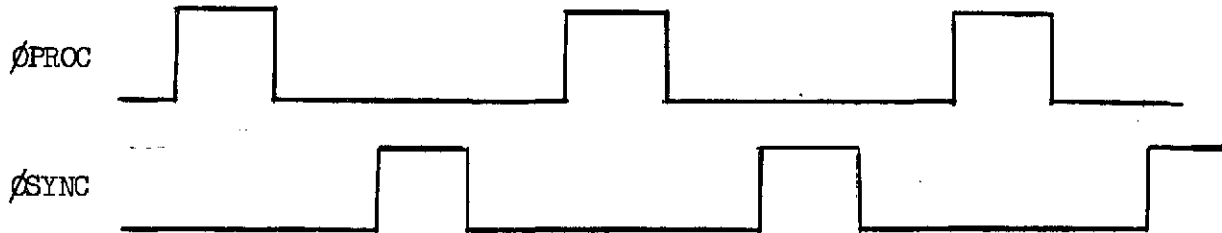
An 11-bit instruction register save (IRS) is used to store the instruction operation code. It is loaded directly from memory during the instruction fetch with the most significant six-bits and the least significant five bits of the instruction word. The remaining six bits exist but are not connected. The most significant or least significant five bits are used for instruction decoding. Bit 13 is the index bit.

A 4-bit page register can be loaded from bits 13-16 of the ACC with a LOAD PAGE instruction or from bits 13-16 of the MOR during an Interrupt, EXIT or RESUME instruction. It is loaded into bits 13-16 of the address register during each instruction fetch and specifies which 4096 word block of memory is to be accessed for an operand fetch.

An 18-bit storage limit register (SL) is used to specify a protected block of memory into which writing is forbidden. The SL is loaded from the adder during an Interrupt, EXIT, or RESUME instruction. Bits 10-18 of the SL specify an upper limit (U). Bits 1-9 of the SL specify a lower limit (L). Bits 8-16 of the ADR specify a block of addresses. If bits 8-16 of the ADR are represented by a number A, then if $L \leq A \leq U$, a write into the address specified by the ADR is permitted. If this condition is not met, the write is not permitted, a "read" occurs and a signal is sent to the SIO which generates an interrupt.

2.2 SYSTEM CLOCKS

The primary system clock is generated in the SIO. In the CPU it is frequency divided in half and two secondary clocks, ϕ PROC and ϕ SYNC, are generated. Their phase relationship is illustrated in Figure A-5.



Clock Phase
Figure A-5

ϕ PROC controls all register transfers in the CPU. ϕ SYNC is primarily used to clear or set R-S latches where it is necessary to extend or shorten a signal by one half of a clock period. ϕ SYNC is used in the I/O control section. CPU registers and control signals generally switch on the trailing edge of ϕ PROC.

2.3 MEMORY INTERFACE

The AOP generates a memory initiate (MINIT) pulse to initiate a memory cycle. During a read operation, the memory responds with a read complete (RC) pulse when the data is available (referencing the access time of the memory). During a write cycle, the CPU puts the MOR on the memory input bus (MIB) and in addition to MINIT, allows a memory bus read/write line (MBRW) to drop to "zero". No read complete signal is generated by the memory during a write operation. During a read cycle the memory output bus (MOB) may be loaded into MOR and IRS. The clocks for both registers are controlled by MINIT which clears the registers by gating the inputs to the flip flops to zero. Then the MOB is gated with RC to d.c. set the flip flops. This d.c. set technique was used to avoid the extra gate delays involved in setting a flip flop with the clocked



input. The MOR uses CPUACK to STROBE memory data in, while the IRS inputs are enabled during a ϕ_1 phase.

2.4 INSTRUCTION PHASING

There are five main phase flip flops used to control CPU activities during the execution of most instructions. These are:

1. ϕ_1 - instruction fetch
2. ϕ_I - indexing
3. ϕ_2 - operand fetch and instruction execution
4. ϕ_3 - primarily shifting operations, also used for longer instructions
5. ϕ_4 - used when ϕ_2 and ϕ_3 do not suffice.

In addition, there are two subphases, A and B, which are used to subdivide the primary phases where necessary. Special purpose phases ϕ_{MSH} , ϕ_{MUL} , ϕ_{MADD} , ϕ_{DSH} , and ϕ_{DSUB} are used to control the MULTIPLY and DIVIDE instructions.

2.5 OPERATION COUNTER

A six-bit operation counter (OPCTR) is used to control the length of shifting instructions, MULTIPLY, DIVIDE, interrupts, etc. It is loaded from the MOR when the shift count is an operand. The shift count is loaded into the IDX at the end of a NORMALIZE instruction.

3.0 I/O OPERATION

Four functions which were originally located in the I/O of the OBP system were redesigned and have been included with the CPU in the AOP. They are:

1. Memory Bus Controller
2. Interrupt holding registers and control
3. DMA request holding registers and control
4. Command load and dump channels



3.1 BUS CONTROLLER

The function of the memory bus controller is to request memory access and break ties between different parts of the system which may request the use of the memory simultaneously. The priorities in decreasing order of importance are:

1. Command channels
2. Cycle steal channels
3. CPU

Whenever a command load or command dump request is received, a latch is set. On the trailing (positive going) edge of the request pulse a signal is generated which sets an enable latch (CMDAEN) on the next ϕ SYNC pulse. On the following ϕ PROC, a command acknowledge (CMDACK) flip flop is set which generates a MINIT and is also sent to the SIO and CPU. If a DMA operation is in progress it is locked out and terminated. CPU memory requests are also locked out. A "Command Clear" is to precede all command requests to clear the entire CPU-I/O.

When a DMA operation begins, an R-S flip flop (LATCH) is set for the duration of the operation (five memory cycles). LATCH locks out all CPU memory requests and can only be reset by the end of the cycle steal operation (\overline{RCT}) or by CMDAEN. LATCH, conditioned by I/O control signals, causes cycle steal acknowledge (CSACK) to be set at the correct times. CSACK is used to generate MINIT and to control data and addresses on the memory buses.

When a CPU memory request (MEMREQ) is generated a CPU acknowledge flip flop (CPUACK) is set unless it is inhibited by either LATCH or CMDAEN. The only time that a CPU request takes precedence over a cycle steal request is during INPUT and OUTPUT instructions. The I/O is not permitted to initiate a cycle steal between the two memory accesses necessary to fetch a device code



and a data word during these instructions. During successive CMA request, one CPU memory cycle is permitted between each full DMA cycle.

The three acknowledge flip flops (CFUACK, CSACK, CMDACK) are "ored" and the resultant signal is used to trigger a monostable multivibrator which produces MINIT. Nodes are available external to the CMMA package which may be capacitively loaded to delay or stretch MINIT to aid in interfacing with memories having different characteristics.

3.2 INTERRUPTS

The AOP is capable of storing up to 16 asynchronous interrupts and servicing them in order of priority. Incoming interrupts are stored in the 16-bit Interrupt Storage Register (ISR). A logic "zero" level is required to set an ISR bit. However, a logic "one" level on the interrupt line is required for any further action. This is to prevent a stuck-at-one or stuck-at-zero faults on an interrupt line from causing continuous interrupts.

A 16-bit Lockout Status Register (LSR) can be used to inhibit any interrupt except interrupt zero. Interrupt zero is used to initiate the machine from a halt state. For each of the remaining 15 interrupts, an interrupt N, where $1 \leq N \leq 15$ can be accepted if ISR bit N is a logic "one", Interrupt line N is a logic "one" and LSR bit N is a logic "zero".

A 16-bit Allowable Interrupt Register (AIR) stores all interrupts which meet the above conditions. The AIR is clocked with ϕ PROC except when an interrupt is being serviced. No new interrupts can be accepted while an interrupt is in progress. They may be stored in the ISR but will not be clocked into the AIR.

The AIR outputs feed a priority decoder which insures that only one interrupt is selected at a time and that interrupts are serviced in order of



priority. Interrupt priorities are in numerical order with interrupt "0" being the highest priority and interrupt "15" the lowest.

Each interrupt priority decoding produces a four bit address code equal to the binary representation of the interrupt number. At the beginning of the interrupt routine, this code is entered into bits 4-7 of the ADR and is used to select a different reserved block of eight memory locations for each interrupt. Four block code bits from the SIO are loaded into bits 13-16 of the ADR to select the 4096 word block of memory desired.

The interrupt routine consists of four write cycles followed by four read cycles. During the first four cycles the LSR, miscellaneous registers (Page, D, carry, overflow), SL and IC are stored in the first four reserved addresses. Then the LSR, miscellaneous, SL and IC are loaded with new values from the second four reserved addresses. The flow of data between the MOR and the other registers is coordinated by the operation counter and the two active phases ϕ_2 and ϕ_3 . The arrangement of the data within the eight word reserved block of memory is indicated in Figure A-6.

Through the use of the SET INTERRUPT OVERRIDE instruction it is possible to prevent all interrupts except interrupt "0" from being serviced. This can be used to allow the interrupting program time to accomplish its job without being interrupted. A RESET INTERRUPT OVERRIDE instruction is available which resets the flip flop which inhibited the interrupts.

Two CPU instructions, RESUME and EXIT, utilize much of the same control logic as the interrupts and perform similar functions. The EXIT instruction is nothing more than a programmed interrupt. It causes the AOP to store the program status information in four locations beginning with 200_g. It then fetches the following four words and loads them into the LSR, miscellaneous



CONTENTS OF INTERRUPT LOCATIONS

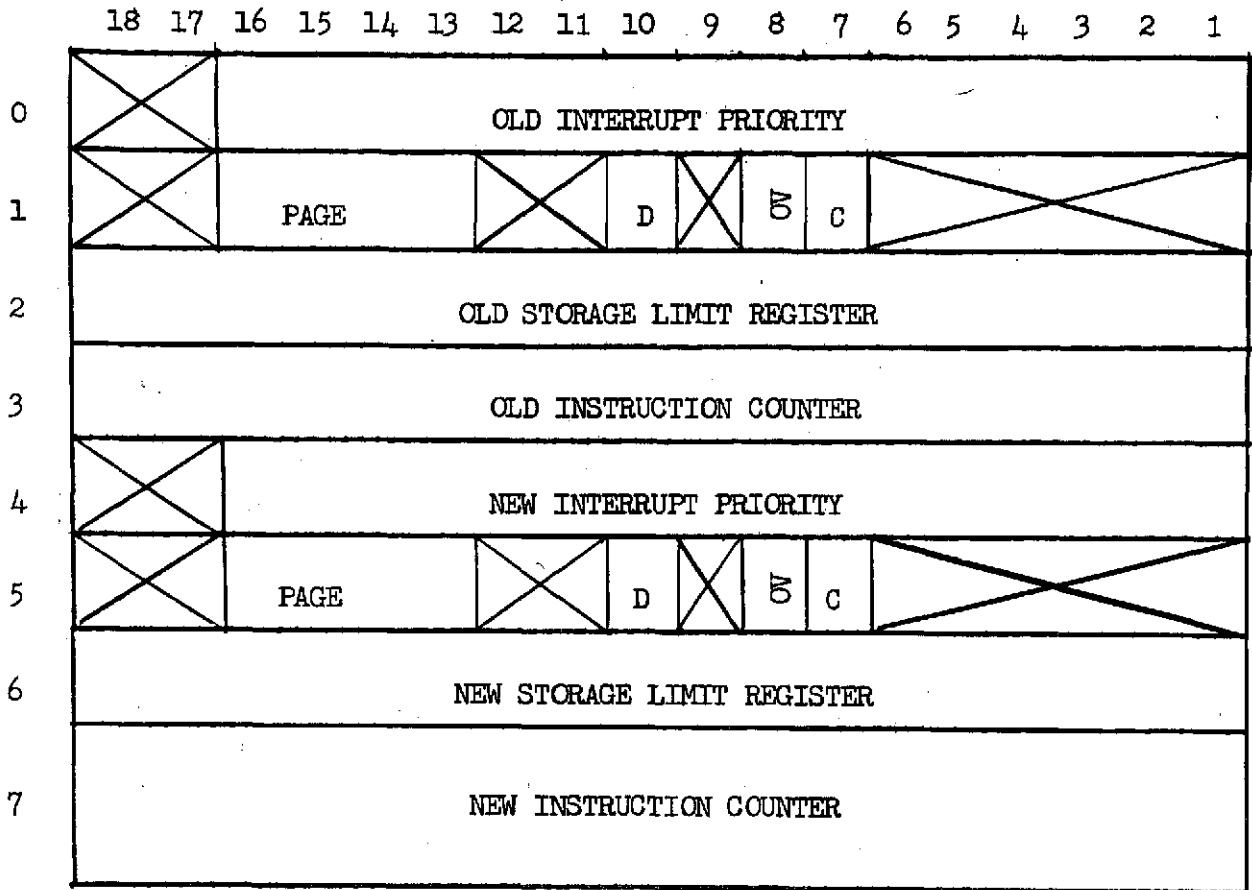


FIGURE A-6 WORD FORMAT IN MEMORY BLOCK

registers, SL and IC. The RESUME instruction is most often used to restore control to the program which was being executed when an interrupt occurred. It goes to an address specified by the instruction word to get a starting address for a block of four memory locations which it loads into the LSR, miscellaneous registers, SL and IC.

If an interrupt occurs during the execution of an instruction it is not honored until the instruction is completed. If several allowable interrupts are stored in the AIR, there must be one CPU instruction executed before each



interrupt is serviced, i.e., it is impossible for the interrupts to prevent a CPU program from being executed, although it may be slowed down.

Master Clearing the system clears the ISR (i.e. erases all interrupts), sets all 16-bits of the LSR (locks out all interrupts except interrupt "0"). Thus, after a master clear only an interrupt "0" can restart the system assuming the test panel is not connected.

After interrupts have been clocked into the AIR and are honored, they are reset. The INTACK signal, which arrives at the beginning of the Interrupt, is used in conjunction with the interrupt being honored to reset the ISR which in turn allows the AIR to be cleared on the next clock. Unless that particular interrupt is requested again, the AIR will remain reset.

3.3 CYCLE STEAL REQUESTS

The AOP has 16 DMA devices which time share a single channel. When a DMA request is honored, the I/O interrupts program execution for 5 memory cycles and then restores operation without affecting any CPU registers. In fact, if the instruction being executed is long and does not require the use of the memory, e.g., MULTIPLY, it may continue without any interruption at all. Cycle steal requests are sent from the I/O as negative going pulses. The hardware which receives them is identical to that for interrupts. A 16-bit Activation Status Register (ASR) can be reset or set under program control to activate or deactivate any I/O channel. If an RSR bit is set ("one" level), the request line has returned to a "one" level, and the corresponding ASR bit is at a "zero" level then the request is entered into the Allowable Request Register (ARR). A 16 level priority circuit selects the highest priority allowable request and initiates a cycle steal routine.

An I/O channel is activated under program control when a block of



data words is to be input to or output from the memory by an I/O device. Before activating a channel, the length of the block and the address of the first word in the block are stored in two reserved memory locations. These addresses (two locations per channel) can be located at the top of any 4096 word block of memory in locations 7740_8 - 7777_8 . The bank is selected by the SIO fixed bits under ground control. The block length is located in the first location and the address in the second. During each cycle steal, the block length is decremented and the address incremented so that the block length always contains the number of words left to be transferred and the address location contains the address of the next word to be transferred.

When an allowable request is detected, a six cycle chain of events occurs. An R-S flip flop, LATCH, is set which locks out the CPU from the memory. LATCH also sets bit 1 of the cycle counter (CT1-6) which is fundamentally a six bit shift register. During CT1 a four bit address code derived from the request priority circuit output is loaded into bits 2-5 of the five bit Cycle Steal Address Register (CSA). Bit one of the CSA is cleared and since bits 6-16 of the CSA do not exist, the memory address bus (MAB) bits in these positions are at "one" levels. The contents of the CSA are used as the address to fetch the block length from memory. When the memory responds, the MOB is strobed into the 16-bit Block Length/Address Register (BLAD). The BLAD is a 16-bit up/down counter.

During CT2 the BLAD is tested for zero and if it is not equal to zero it is decremented. If it is equal to zero the cycle steal routine is terminated and that ASR bit is set.

During CT3 the BLAD is again tested for zero. If it now equals zero the "block length = 0" interrupt (BLINT) is generated and sent to the SIO.



The BLAD is stored back in the same memory location from which it was fetched. Bit 1 of the CSA is then set, thereby increasing the address by one.

During CT4 the address of the data word to be transferred is fetched and loaded into the BLAD.

During CT5 the data word is transferred either in or out of memory. The contents of the BLAD are used as a memory address. CSACK is gated with the request priority to produce a Special I/O Acknowledge (SPACK) which signals the SIO to either gate a data word onto the MIB or to take one off the MOB when read complete is generated. If the channel being serviced is in the range 0-7, a write signal is sent to the memory. Otherwise, the channel is an output channel and a read cycle results. At the end of the CT4 the BLAD is incremented by one.

During CT6 the CSA is once again used as the memory address and the contents of the BLAD are written into the same location from which they were fetched in CT4. LATCH is reset and control is returned to the CPU.

The request for the activation of a particular device results in the RSR being set which in turn allowed the ARR to be clocked "set" on the next clock. After that particular request is honored or the block length has reached a "zero", the RSR and ARR are reset. The REQR signal is used in conjunction with the particular device being honored to reset the RSR device. The next clock in turn resets the ARR. REQR is generated when a block length of zero is encountered during a cycle steal request or at the end of each cycle steal request. The request must be renewed to honor that device again.

3.4 COMMAND CHANNELS

Two command actuated channels, Command Load and Command Dump provide



the capability of rapidly loading or dumping the contents of the entire memory.

The existence of these hardwired channels eliminates any requirement for a bootstrap program of any sort to reload memory in the event of a failure. The system can be cleared and reloaded from the ground without difficulty and without extra hardware such as a read only section of memory to store a bootstrap load program. It is assumed that a command operation will always be preceded by a command clear of the system. This is necessary to insure that the BLAD contents are zero initially. For both command operations, the BLAD provides bits 1-12 of the address and is incremented by one with each memory cycle. Four block code bits from the SIO select one 4096 word block of memory. Command load requests receive memory write cycles whereas Command Dump requests generate memory read cycles. The requests set latches when the request lines go to "zero" levels. When they return to a "one" level the sequence of events leading to the memory cycle begin. As with the interrupts and cycle steal requests, this prevents a stuck-at-zero or stuck-at-one command request line from generating continuous command memory cycles. The latches are reset at the end of the memory cycle.

Command Load requests are accepted indefinitely. However, Command Dump requests are inhibited after bit 15 of the BLAD is set. Bit 15 of the BLAD sets a latch which lockout command dump requests after four passes through the selected bank have been completed. This latch may only be reset by master clearing the system.

4.0 AOP INSTRUCTION SET

The 55 instructions available in the AOP are determined by decoding a five (5) bit major op-code field or a five bit minor field. All instructions



consist of an instruction fetch phase followed by an operational phase. This operational phase, in major op code instructions includes an operand fetch cycle except for Load Accumulator with Effective Address (LDL = 40).

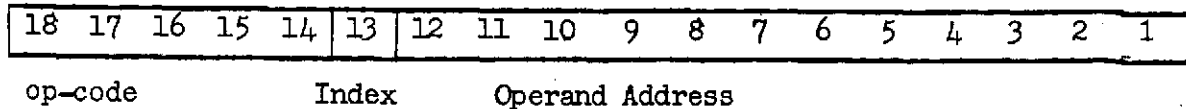
During the instruction fetch phase, an op code field is fetched from memory. This field is subdivided into the most significant five bits (MS5B) and the least significant five bits (LS5B) of the memory word. The MS5B define the "major Op-codes" which include the operand fetch cycle. The least significant twelve (12) bits of the major op-code field are then used, along with a four bit page register, to form the effective address for the operand fetch. When the MS5B are all zeros, then the LS5B are used to decode the "minor op-codes" which do not have operand fetch cycles. The "EXIT" instruction contains additional memory cycles as a "minor op-code instruction.

Indexing of instructions is accomplished when bit 13 of the instruction fetch word is a "1". Two clock cycles are required for indexing in which the index register (IDX) (18 bits) is added to the effective address (ADR) (18 bits) following the instruction fetch phase. Only the least significant 16 bits of effective address are used for addressing. During a "minor op-code" instruction, no useful result is obtained from indexing. Figure A-7 illustrates the format of the instruction words.

The following list of instructions is arranged in alphabetical order. Execution times are given in machine cycles including instruction fetch but excluding indexing. Indexing adds two machine cycles to any instruction. A machine cycle is equal to one period of the processor clock (ϕ PROC).



Major Op-Code (Memory Access)



Minor Op-Code

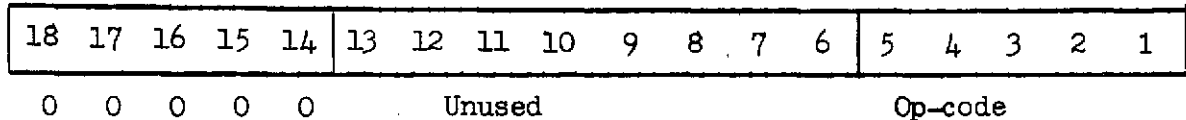


FIGURE A-7

INSTRUCTION WORD FORMATMajor Op-Codes

Add Memory to Accumulator

ADD O4

The contents of storage at the effective address are added to the contents of the accumulator and the sum is retained in the accumulator. If a carry out of bit 17 of the adder occurs, the carry register is set to one. Otherwise, the carry register is reset to zero. Overflow can occur when two numbers of the same sign are added. If bit 18 of the sum differs from bit 18 of the operands, the overflow register is set to one. Otherwise it is unchanged. The changed sign bit of the sum is retained in the accumulator.

Register altered: Accumulator

Carry register (conditionally)

Overflow register (conditionally)

Timing: 4 cycles



Add Memory to Index

ADX 02

The contents of storage at the effective address are added to the contents of the index register. The 18-bit sum is retained in the index register.

Register altered: Index register

Timing: 4 cycles

Branch Conditionally

BRC 42

If the contents of the decision register are zero, then the next instruction is fetched from the address specified by the present contents of the IC. If the content of the decision register is one, then the contents of storage at the effective address are placed in the IC and execution proceeds from the new address specified by the instruction counter. The decision register is reset to zero.

Registers altered: Decision register (always reset to zero)

Timing: 4 cycles

Branch and Mark Place

BRM 06

The contents of the instruction counter plus one are stored at the effective address unless that address is protected by the storage limit register. If storage is protected, no write into memory occurs and a signal (Outlimit) is sent to the SIO. The contents of one location greater than the effective address is placed in the instruction counter and execution proceeds from the new address specified by the instruction counter.

Registers altered: None

Timing: 8 cycles



Branch Unconditionally

BRU 62

The contents of storage at the effective address are placed in the IC and execution proceeds from the new address specified by the IC.

Registers altered: None

Timing: 4 cycles

Cyclic Shift Accumulator

CYC 34

The least significant 6 bits of the contents of storage at the effective address are used as a two's complement shift count. If the count is negative, then the contents of the accumulator are shifted cyclically right the number of bits specified by the count, with bits leaving bit 1 entering bit 18. If the count is positive, then the contents of the accumulator are shifted left the number of bits specified by the count with bits leaving bit 18 entering the bit 1.

Registers altered: Accumulator

Timing: 5 cycles + 1 cycle per position shifted.

Double Cyclic Shift

DCY 56

The low-order 6 bits of the contents of storage at the effective address are used as a two's complement shift count. If the count is negative, then the contents of the accumulator and extended accumulator are shifted cyclically right the number of positions specified by the count with bit 1 of the extended accumulator entering bit 18 of the accumulator and bit 1 of the accumulator entering bit 18 of the extended accumulator. If the count is positive, then the direction of the above process is reversed.

Registers altered: Extended accumulator and Accumulator

Timing: 5 cycles + 1 cycle per position shifted



Double Shift

DSH 36

The low-order 6 bits of the contents of storage at the effective address are used as a two's complement shift count. The accumulator and the extended accumulator are shifted together. If the count is negative, then both registers are shifted right the number of bits specified by the count. Bit 18 of the accumulator is shifted into vacated bits to the right. Bit 18 of both registers is unchanged. Bit 1 of the accumulator is shifted into bit 17 of the extended accumulator.

If the count is positive both registers are shifted left the number of bits specified by the count. Zeros are shifted into bit 1 of the extended accumulator. Bit 17 of the extended accumulator is shifted into bit 1 of the accumulator. Bit 18 of the extended accumulator is unchanged. If bit 18 of the accumulator changes during the instruction the overflow register is set to one.

Timing: 5 cycles + 1 cycle per position shifted

Divide

DIV 64

The accumulator and extended accumulator form the dividend that is divided by the contents of storage at the effective address. The signed remainder is retained in the accumulator and the signed quotient is retained in the extended accumulator. The divisor and dividend must be positive and the dividend must be less than or equal to the divisor. Otherwise the results are unpredictable. The quotient and remainder are positive and the remainder has a magnitude less than the divisor. If the dividend is not less than or equal to the divisor the resulting overflow is not detected.



Registers altered: Accumulator

Extended accumulator

Timing: 56-58 cycles

Exclusive OR

EOR 70

The contents of storage at the effective address are exclusive ORed with the contents of the accumulator. The result is retained in the accumulator. The 18 bits of the result are computed independently with a one occurring in a bit position of the result if either the accumulator or storage, but not both; contain a one in that bit position.

Registers altered: Accumulator

Timing: 4 cycles

Input From

IPF 76

Two addresses are accessed: n (the effective address) and $n + 1$. The contents of bits 1-4 of n are decoded to select an input device (0-7). One word is transferred from the selected device and stored in $n + 1$.

Registers altered: None

Timing: 6 cycles

Load Accumulator

LDA 20

The accumulator is loaded with the contents of the effective address.

Registers altered: Accumulator

Timing: 4 cycles

Load Extended Accumulator

LDE 52

The extended accumulator is loaded with the contents of the effective address.

Registers altered: Extended accumulator

Timing: 4 cycles



Load Indirect

LDI 12

The 16 LSB's of memory at the effective address are treated as a new effective address. If bit 18 of the memory word is a one, the contents of the index register are added to the new effective address. Otherwise it remains unchanged. The contents of memory at the new effective address are placed in the accumulator.

Registers altered: Accumulator

Timing: 6 cycles

Load Location

LDL 40

The accumulator is loaded with the effective address.

Registers altered: Accumulator

Timing: 4 cycles

Load Index

LDX 54

The index register is loaded with the contents of the effective address.

Registers altered: Index register

Timing: 4 cycles

Logical AND

ETR 30

The contents of storage at the effective address are logically ANDed with the contents of the accumulator. The result is retained in the accumulator. The 18 bits of the result are computed independently with a one occurring in a bit position of the result only if the accumulator and storage both contained a one in that bit position.

Registers altered: Accumulator

Timing: 4 cycles

Logical OR

MRG 50

The contents of storage at the effective address are logically ORed



with the contents of the accumulator. The result is retained in the accumulator. The 18 bits of the result are computed independently with a one occurring in a bit position of the result if either the accumulator or storage contained a one in that position.

Registers altered: Accumulator

Timing: 4 cycles

Multiply

MUL 44

The contents of storage at the effective address are multiplied by the contents of the accumulator. The high-order 17 bits and sign of the product are retained in the accumulator. The low-order 17 bits and sign of the product are retained in the extended accumulator.

Registers altered: Accumulator

Extended accumulator

Timing: 24 to 42 cycles

Output TO

OPT 16

Two addresses are accessed: n (the effective address) and $n + 1$. The contents of bits 1-4 of n are decoded to select an output device (8-15). One word is transferred from location $n + 1$ to the selected device. The activation status register in the AOP is device 10.

Registers altered: None

Timing: 6 cycles

Resume from Interrupt

TIM 72

The contents of storage at the effective address is used as the starting address of a 4-word save area. This instruction restores the registers that were saved in these locations (i.e., by the occurrence of an interrupt). Upon completion, execution proceeds normally at the new value in the instruction counter.



Registers altered: Lockout status register
Storage limit register
Page register
Overflow register
Carry register
Decision register
Instruction counter

Timing: 22 cycles

Shift

SHF 14

The low-order 6 bits of the contents of storage at the effective address are used as a two's complement shift count. If the count is negative, then the accumulator is shifted right the number of positions specified by the count, with the contents of the accumulator sign (bit 18) replacing vacated positions to the right. If the count is positive, then the accumulator is shifted left the number of positions specified by the count with zeros filling vacated positions to the left of bit 1. The overflow register is set to one if the sign (bit 18) of the accumulator is changed during the shift. If the sign bit of the accumulator does not change the overflow register is unaffected by the instruction.

Registers altered: Accumulator

Overflow register (conditionally)

Timing: 5 cycles + 1 cycle per position shifted

Store Accumulator

STA 60

The contents of the accumulator are stored at the effective address unless that address is protected by the storage limit registers. If storage is protected, no write into memory occurs and a signal (out-limit) is sent to the SIO.

Registers altered: None

Timing: 6 cycles



Store Extended Accumulator

STE 10

The contents of the extended accumulator are stored at the effective address unless that address is protected by the storage limit registers. If storage is protected, no write into memory occurs and a signal (out-limit) is sent to the SIO.

Registers altered: None

Timing: 6 cycles

Store Accumulator Indirect

STI 32

The contents of memory at the effective address are loaded into the MOR. The contents of the MOR are loaded into the ADR through the adder. If bit 18 of the MOR is one, the contents of the IDX are added to it during the transfer. Otherwise, zero is added to the MORE. The new ADR used as an effective address remains unchanged. The contents of the accumulator are stored in memory at the new effective address unless that location is protected by the storage limit register. If storage is protected, no write into memory occurs and a signal (outlimit) is sent to the SIO

Registers altered: None

Timing: 8 cycles

Store Index

STX 74

The contents of the index register are stored at the effective address unless that address is protected by the storage limit registers. If storage is protected, no write into memory occurs and a signal (outlimit) is sent to the SIO.

Registers altered: None

Timing: 6 cycles



Subtract

SUB 24

The contents of storage at the effective address are subtracted from the contents of the accumulator and the result remains in the accumulator. Subtraction is performed by adding the two's complement of the contents of storage to the accumulator (a carry is forced into the low order stage of the adder). If a carry occurs out of the 17th position, the carry register is set to one. Otherwise it is reset to zero. Overflow can occur when two numbers of unlike sign are subtracted. Overflow causes the overflow register to be set to one and the 18th bit of the difference is retained in the sign position.

Registers altered: Accumulator

Carry register (conditionally)

Overflow register (conditionally)

Timing: 4 cycles

Test Accumulator Equal

TAE 46

If the contents of the accumulator are equal to the contents of storage at the effective address then the decision register is set to one. Otherwise it is unchanged.

Registers altered: Decision register (conditionally)

Timing: 4 cycles

Test Accumulator Greater

TAG 66

If the contents of the accumulator are greater than the contents of storage at the effective address, the decision register is set to one. Otherwise, it is unchanged.

Registers altered: Decision register (conditionally)

Timing: 4 cycles



Test Accumulator Less

TAL 26

If the contents of the accumulator are less than the contents of storage at the effective address, then the decision register is set to one.

Otherwise, it is unchanged.

Registers altered: Decision register (conditionally)

Timing: 4 cycles

Test Index Less

TXLE 22

If the contents of the index register are less than or equal to the contents of storage at the effective address, then the decision register is set to one. Otherwise, it is unchanged.

Registers altered: Decision register (conditionally)

Timing: 4 cycles

Minor Op-Codes

Add Carry

ADC 06

The content of the carry register is added to the contents of the accumulator and the 18 bit sum is retained in the accumulator. If a carry occurs at the output of the 17th bit of the adder, the carry register is set to one. Otherwise, the carry register is reset to zero. Overflow occurs if bit 18 of the sum differs from bit 18 of the original contents of the accumulator. If overflow occurs, the overflow register is set to one. Otherwise, it is unchanged.

Timing: 4 cycles

Complement Accumulator

CMP 10

The one's complement of the contents of the accumulator is loaded into the accumulator.



Registers altered: Accumulator

Timing: 6 cycles

Complement Decision

CPD 17

The content of the decision register is complemented.

Registers altered: Decision register

Timing: 3 cycles

Exchange Accumulator and Extension

AEA 26

The contents of the accumulator and extended accumulator are interchanged.

Registers altered: Accumulator

Extended accumulator

Timing: 8 cycles

Exchange Accumulator and Index

ACX 25

The contents of the accumulator and index registers are interchanged.

Registers altered: Accumulator

Index register

Timing: 8 cycles

Exchange Extension and Index

EAX 27

The contents of the extended accumulator and index register are interchanged.

Registers altered: Extended accumulator

Index register

Timing: 8 cycles

EXIT

EXIT 16

This instruction initiates a programmed interrupt which uses locations octal 200 through 207. The status of registers is saved in 200-203 and these registers are then loaded from locations 204-207. Upon



completion, execution proceeds normally at the new value in the instruction counter.

Registers altered: Lockout status register
Storage limit register
Page register
Overflow register
Carry register
Decision register
Instruction counter
Address register

Timing: 36 cycles

HALT

HLT 00

The processor stops indefinitely. An interrupt signal must be supplied from an external source to start the processor. The instruction counter is updated.

Registers altered: None

Timing: 3 cycles before an interrupt will be honored.

Load Extended Accumulator with Decision

LDD 13

The contents of the accumulator and extended accumulator are shifted left one position. The sign (bit 18) of the extended accumulator is not shifted. Bit 17 of the extended accumulator is shifted into bit 1 of the accumulator and the vacated, low-order position of the extended accumulator is filled with the contents of the decision register. The overflow register is not altered.

Registers altered: Accumulator

Extended accumulator

Timing: 3 cycles

Load Page

LDP 12

The contents of bits 13 through 16 of the accumulator are placed in the page register.



Registers altered: Page register

Timing: 3 cycles

Negate Accumulator

NEG 04

The contents of the accumulator are replaced by its two's complement. Negating all zeros yields a result of zero and sets the carry register to one. Negating the number that has zeros in all bit positions except the sign yields the same number as a result and sets both the carry register and the overflow register to one. Other than these two special cases, the carry register is reset to zero.

Registers altered: Accumulator

Carry register (conditionally)

Overflow register (conditionally)

Timing: 6 cycles

No Operation

NOP 02

No operation is performed other than the automatic incrementing of the instruction counter.

Register altered: None

Timing: 3 cycles

Normalize

NORM 14

The contents of the accumulator and extended accumulator are shifted left until the 17th and 18th bits of the accumulator are different. The sign bit of the extended accumulator is not shifted. Bit 17 of the extended accumulator is shifted into bit 1 of the accumulator. Bit 18 of the extended accumulator is not changed. Zeros fill the positions vacated on the right. A count of the number of positions shifted is retained in bits 1-6 of the index register. If the contents of the accumulator and bit positions 1 through 17 of the extended accumulator



are zero, then the index register is set to zero.

Registers altered: Accumulator
Extended Accumulator
Index register

Timing: 5 cycles + 1 cycle per position shifted.

Reset Decision

RED 23

The decision register is reset to zero.

Registers altered: Decision register

Timing: 3 cycles

Reset Interrupt Override

RIO 24

The interrupt override register is reset to zero.

Registers altered: Interrupt override register.

Timing: 3 cycles

Reset Overflow

ROV 07

The overflow register is reset to zero.

Registers altered: Overflow register

Timing: 3 cycles

Reverse Accumulator

FLP 22

The contents of the accumulator are reversed. The $(19-n)^{\text{th}}$ and n^{th} bits are exchanged for $n = 1, 2, \dots, 9$.

Registers altered: Accumulator

Timing: 3 cycles

Set Interrupt Override

SIO 20

The interrupt override register is set to one, thereby locking out all interrupts except interrupt zero.

Registers altered: Interrupt override register

Timing: 3 cycles



Test Accumulator Positive

TAP 03

If the sign position, bit 18, of the accumulator contains a zero, then the contents of the decision register are set to one. Otherwise, it is unchanged.

Registers altered: Decision register (conditionally)

Timing: 3 cycles

Test Accumulator Zero

TAZ 21

If the contents of the accumulator is equal to zero, then the contents of the decision register are set to one. Otherwise, it is unchanged.

Registers altered: Decision register (conditionally)

Timing: 4 cycles

Test and Increment Extended Accumulator

TIE 15

The contents of the decision register are reset to zero. Then the contents of the extended accumulator are tested for zero. If it is non-zero, the contents of the decision register are set to one and the extended accumulator is incremented by one. If the contents of the extended accumulator are zero, the decision register and extended accumulator remain unchanged.

Registers altered: Decision register (conditionally)

Extended accumulator (conditionally)

Timing: 6 cycles

Test and Increment Index

TIX 11

The contents of the decision register is reset to zero. Then the contents of the index register are tested for zero. If it is non-zero, the contents of the decision register are set to one and the contents of the index registers are incremented by one. If the contents of the index register are zero, the decision and index registers



remain unchanged.

Registers altered: Decision register (conditionally)

Index register (conditionally)

Timing: 6 cycles

Test Accumulator for Odd Parity

TOP 05

If the number of ones in the 18 bit accumulator is odd, then the contents of the decision register is set to one. Otherwise, it is unchanged.

Registers altered: Decision register (conditionally)

Timing: 22 cycles

Test Overflow

TOV 01

If the contents of the overflow register are one, then the decision register is set to one. Otherwise, it is unchanged. The overflow register is reset to zero.

Registers altered: Decision register (conditionally)

Overflow register

Timing: 3 cycles